

$$1) \quad x(k+1) = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} x(k) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(k)$$

$$u(k) = -Lx(k) + Kr(k) = -[l_1, l_2]x(k) + Kr(k)$$

$$\Rightarrow x(k+1) = \underbrace{\begin{bmatrix} 1-l_1 & -l_2 \\ 1 & 1 \end{bmatrix}}_{A_c} x(k) + \underbrace{\begin{bmatrix} K \\ 0 \end{bmatrix}}_{B_r} r(k)$$

a) Poles  $\equiv$  eigenvalues of  $A_c$

$$\begin{aligned} \det(\lambda I - A_c) &= \det \begin{bmatrix} \lambda - 1 + l_1 & l_2 \\ -1 & \lambda - 1 \end{bmatrix} \\ &= (\lambda - 1)(\lambda - 1 - l_1) + l_2 \\ &= \lambda^2 + \lambda(l_1 - 2) + l_2 - l_1 + 1 \\ &= (\lambda - 0.9)^2 \\ &= \lambda^2 - 1.8\lambda + 0.81 \end{aligned}$$

$$\Rightarrow \underline{l_1 = 0.2} \quad \text{and} \quad l_2 = 0.81 + l_1 - 1 = \underline{0.01}$$

$$b) \quad \begin{aligned} x(k+1) &= A_c x(k) + B_r r(k) \\ y(k) &= C x(k) \end{aligned}$$

$$\text{Stable \& SS} \Rightarrow x(k+1) = x(k) = \bar{x}$$

$$(I - A_c)\bar{x} = B_r \bar{r}$$

$$\Rightarrow \bar{y} = C(I - A_c)^{-1} B_r \bar{r}$$

$$= [0 \ 1] \begin{bmatrix} l_1 & l_2 \\ -1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} K \\ 0 \end{bmatrix} \bar{r}$$

$$= [0 \ 1] \frac{1}{l_2} \begin{bmatrix} 0 & -l_2 \\ 1 & l_1 \end{bmatrix} \begin{bmatrix} K \\ 0 \end{bmatrix} \bar{r}$$

$$= 1 \Rightarrow 100K = 1 \Rightarrow \underline{\underline{K = 0.01}}$$

2) If reachable then poles can be placed arbitrarily

Controllability matrix

$$C = [B \ AB \ A^2B] = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

full rank  $\Rightarrow$  reachable

$$3) \quad G(p) = \begin{bmatrix} 0 & \frac{2}{p+1} \\ 0 & \frac{1}{p} \\ \frac{1}{p+1} & \frac{2}{p+1} \end{bmatrix}$$

a) Largest minors (order 2)

$$0, \quad -\frac{1}{p(p+1)}, \quad \frac{-2}{(p+1)^2}$$

Order 1 = elements of  $G$

The pole polynomial is the least common divisor of all minors, i.e.

$$p(p) = p(p+1)^2$$

$\therefore$  3 poles:  $0, -1, -1$

The minors are consequently

$$\frac{0}{p(p)} \quad , \quad \frac{p+1}{p(p)} \quad , \quad \frac{2p}{p(p)}$$

Zero polynomial is the largest common divisor, which is 1 in this case

$\Rightarrow$  No zeros

b)  $y_1 = \frac{2}{p+1} u_2$

$$y_2 = \frac{1}{p} u_2$$

$$y_3 = \frac{1}{p+1} u_1 + \frac{2}{p+1} u_2$$

Let, for example

$$x_1 = \frac{1}{p+1} u_2, \quad x_2 = \frac{1}{p} u_2, \quad x_3 = \frac{1}{p+1} u_1$$

$$\Rightarrow \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 & 1 \\ 0 & 1 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

$$\begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = \begin{bmatrix} 2 & 0 & 0 \\ 0 & 1 & 0 \\ 2 & 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

Minimal since 3 states and 3 poles  
(observable and controllable equivalently)

$$4) a) \quad y(k) = \frac{1}{q-0.5} u(k) \Rightarrow y(k+1) - 0.5y(k) = u(k)$$

$$\therefore x(k+1) = 0.5x(k) + u(k)$$

$$y(k) = x(k)$$

$$b) \quad x_I(k) = \frac{1}{q-1} (r(k) - y(k))$$

$$\Rightarrow x_I(k+1) - x_I(k) = r(k) - x(k)$$

$$\therefore \underbrace{\begin{bmatrix} x(k+1) \\ x_I(k+1) \end{bmatrix}}_{x_e(k+1)} = \underbrace{\begin{bmatrix} 0.5 & 0 \\ -1 & 1 \end{bmatrix}}_{A_e} \underbrace{\begin{bmatrix} x(k) \\ x_I(k) \end{bmatrix}}_{x_e(k)} + \underbrace{\begin{bmatrix} 1 \\ 0 \end{bmatrix}}_{B_e} u(k) + \underbrace{\begin{bmatrix} 0 \\ 1 \end{bmatrix}}_{K_r} r(k)$$

$$y(k) = \underbrace{[1 \quad 0]}_{C_e} x_e(k)$$

c) Optimal control law (Assume  $r=0$  and  $L_e = [l_1 \quad l_2]$ )

$$\begin{aligned} u(k) &= -L_e x_e(k) \\ &= -l_1 x(k) - l_2 x_I(k) \\ &= -l_1 y(k) + l_2 \frac{1}{q-1} y(k) \end{aligned}$$

Block scheme gives

$$u(k) = -\left(K_p + K_I \frac{1}{q-1}\right) y(k)$$

$$\therefore K_p = l_1 \quad \text{and} \quad K_I = -l_2$$

d) The obtained LQ controller guarantees a stable system

$$5) \quad G(s) = \frac{1}{s} e^{-0.6s}$$

Time delay  $T_d = 0.6$ , Sampling time  $h = 1$

We recognise this process as a pure integration ( $1/s$ ) with a time delay, i.e.

$$a) \quad \dot{x}(t) = ax(t) + bu^*(t), \quad u^*(t) = u(t - 0.6)$$

$$y(t) = cx(t)$$

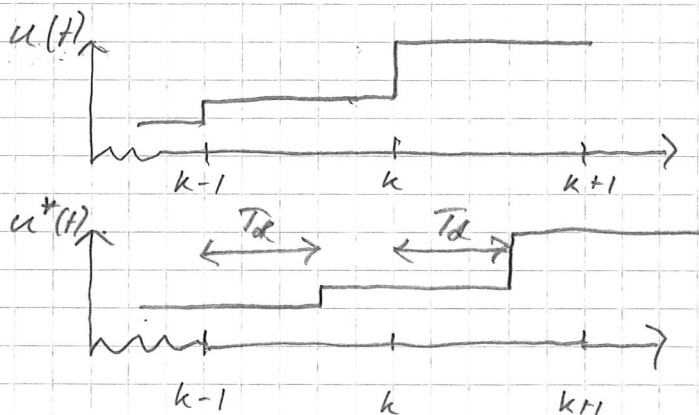
with  $a = 0$ ,  $b = 1$ ,  $c = 1$

b) Given the state  $x_0$  at time  $t_0$  the analytical solution is

$$x(t) = e^{a(t-t_0)} x(t_0) + \int_{t_0}^t e^{a(t-\tau)} B u^*(\tau) d\tau$$

Let  $t_0 = kh = k$  and  $t = (k+1)h = k+1$ .

Study input signal on the interval



Consequently, splitting the interval into two

$$x(k+1) = x(k) + \int_k^{k+T_d} B u(k-1) d\tau + \int_{k+T_d}^{k+1} B u(k) d\tau$$

$$x(k+1) = x(k) + T_d u(k-1) + (1-T_d)u(k)$$

Inroduce delayed input state

$$x_d(k) = u(k-1)$$

Then with  $T_d = 0.6$

$$\begin{bmatrix} x(k+1) \\ x_d(k+1) \end{bmatrix} = \begin{bmatrix} 1 & 0.6 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(k) \\ x_d(k) \end{bmatrix} + \begin{bmatrix} 0.4 \\ 1 \end{bmatrix} u(k)$$

$$b) \quad x(k+1) = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}}_A x(k) + \underbrace{\begin{bmatrix} 0 \\ 1 \end{bmatrix}}_B u(k) + v_1(k)$$

$$y(k) = \underbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}_C x(k) + v_2(k)$$

$$v_1 \sim \text{WGN} \left( \begin{bmatrix} 0 \\ 0 \end{bmatrix}, \begin{bmatrix} 1 & 0 \\ 0 & 2 \end{bmatrix} \right) \Rightarrow R_1 = \begin{bmatrix} 1 & 0 \\ 0 & 2 \end{bmatrix}$$

$$v_2 \sim \text{WGN}(0, 1) \Rightarrow R_2 = 1$$

$$E\{v_1, v_2\} = \begin{bmatrix} 0 \\ 0 \end{bmatrix} \Rightarrow R_{12} = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

$$a) \quad \mathcal{O} = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

$\text{rank}(\mathcal{O}) = 2 \Rightarrow \text{observable}$

b) Standard Kalman (predictor) in SS:

$$\hat{x}_p(k+1) = A\hat{x}_p(k) + Bu(k) + K(y(k) - C\hat{x}_p(k))$$

$$\text{where } \hat{x}_p(k) = \hat{x}(k|k-1)$$

$\text{Var}\{\hat{x}_p(k)\} = P$  where  $P$  is given by

$$P = APAT + R_1 - K(CPA^T + R_{12}^T)$$

$$K = (APC^T + R_{12}) (R_2 + CPC^T)^{-1}$$

$$= \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} P_{11} & P_{12} \\ P_{12} & P_{22} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \left( 1 + \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} P_{11} & P_{12} \\ P_{12} & P_{22} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \right)^{-1}$$

$$= \begin{bmatrix} P_{12} \\ 0 \end{bmatrix} \frac{1}{1 + P_{11}}$$

$$\begin{bmatrix} p_{11} & p_{12} \\ p_{21} & p_{22} \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix}}_{\begin{bmatrix} p_{22} & 0 \\ 0 & 0 \end{bmatrix}} + \begin{bmatrix} 1 & 0 \\ 0 & 2 \end{bmatrix} - \frac{1}{1+p_{11}} \begin{bmatrix} p_{12} \\ 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix}$$

$$\begin{bmatrix} p_{12}^2 & 0 \\ 0 & 0 \end{bmatrix}$$

$$\Rightarrow \left. \begin{aligned} p_{11} &= p_{22} + 1 - \frac{p_{12}^2}{1+p_{11}} \\ p_{12} &= 0 \\ p_{22} &= 2 \end{aligned} \right\} p = \begin{bmatrix} 3 & 0 \\ 0 & 2 \end{bmatrix} \Rightarrow K = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

c) Filter case

(1)  $\hat{x}_f(k+1) = \hat{x}_p(k+1) + K_f (y(k) - C\hat{x}_p(k))$

where  $\hat{x}_f(k) = \hat{x}(k|k)$  and

(2)  $\hat{x}_p(k+1) = A\hat{x}_p(k) + Bu(k)$

$$\Rightarrow K_f = PC^T(CPC^T + R_2)^{-1}$$

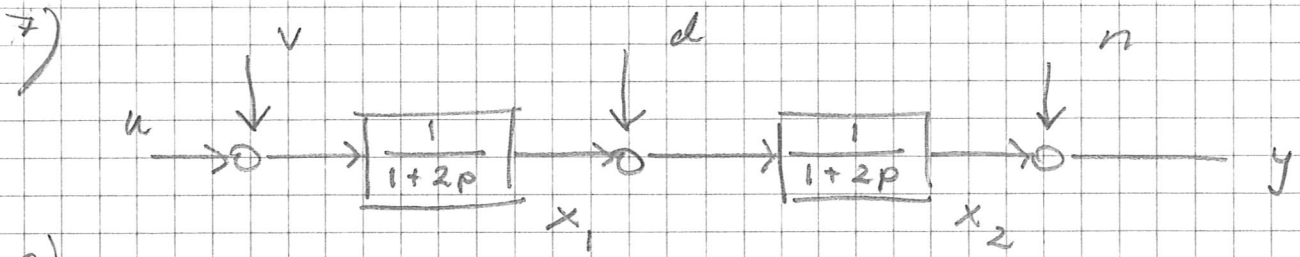
$$= \begin{bmatrix} 3 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \frac{1}{1+p_{11}} = \begin{bmatrix} 3/4 \\ 0 \end{bmatrix}$$

$$\text{Var}\{\hat{x}_f\} = P_f = P - PC^T(CPC^T + R_2)^{-1}CP$$

$$= \begin{bmatrix} 3 & 0 \\ 0 & 2 \end{bmatrix} - \begin{bmatrix} 3/4 \\ 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} 3 & 0 \\ 0 & 2 \end{bmatrix} = \begin{bmatrix} 0.75 & 0 \\ 0 & 2 \end{bmatrix}$$

Var $\{\hat{x}_2\}$  unchanged but Var $\{\hat{x}_1\}$  reduced by 75%





a) With  $x_1$  and  $x_2$  as above

$$x_1 = \frac{1}{1+2p} (u+v) \Leftrightarrow \dot{x}_1 = -\frac{1}{2}x_1 + \frac{1}{2}u + \frac{1}{2}v$$

$$x_2 = \frac{1}{1+2p} (x_1+d) \Leftrightarrow \dot{x}_2 = -\frac{1}{2}x_2 + \frac{1}{2}x_1 + \frac{1}{2}d$$

Spectral factorization

$$\Phi_{\downarrow d}(\omega) = \frac{1}{\omega^2 + 1} \Leftrightarrow d(t) = \frac{1}{p+1} v(t)$$

if  $v(t)$  white noise with unit intensity

$$\text{Let } d = x_3 \Rightarrow \dot{x}_3 = -x_3 + v$$

$$\therefore \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -0.5 & 0 & 0 \\ 0.5 & -0.5 & 0.5 \\ 0 & 0 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0.5 \\ 0 \\ 0 \end{bmatrix} u + \begin{bmatrix} 0.5 & 0 \\ 0 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} v \\ v \end{bmatrix}$$

$$y = \begin{bmatrix} 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + n(t)$$

where  $e = \begin{bmatrix} v \\ v \end{bmatrix}$  have intensity  $\begin{bmatrix} \Phi_v & 0 \\ 0 & \Phi_v \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$

$$b) \Phi_n = \frac{\omega^2 + 4}{\omega^2 + 9}$$

Spectral factorization again gives

$$n = \frac{p+2}{p+3} v_2 \quad , \quad v_2 \text{ white noise with unit intensity}$$

$$n = \frac{p+3}{p+3} v_2 - \frac{1}{p+3} v_2 = v_2 - x_4$$

$$\Rightarrow \dot{x}_4 = -3x_4 + v_2$$

$$\therefore \dot{x} = \begin{bmatrix} -0.5 & 0 & 0 & 0 \\ 0.5 & -0.5 & 0.5 & 0 \\ 0 & 0 & -1 & 0 \\ 0 & 0 & 0 & -3 \end{bmatrix} x + \begin{bmatrix} 0.5 \\ 0 \\ 0 \\ 0 \end{bmatrix} u + \begin{bmatrix} 0.5 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \underbrace{\begin{bmatrix} v_1 \\ v_2 \\ v_2 \end{bmatrix}}_{v_1}$$

$$y = [0 \ 1 \ 0 \ -1] x + v_2$$

c)  $v_1, d, n$  independent  $\Rightarrow$

$v_1, v_2, v_2$  — " — "

$$\Rightarrow R_1 = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$R_{12} = \begin{bmatrix} \Phi_{v_1 v_2} \\ \Phi_{v_2 v_2} \\ \Phi_{v_2 v_2} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} \quad , \quad R_2 = \Phi_{v_2} = 1$$

$$\therefore R = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 1 \end{bmatrix}$$