SSY080

Transformer, Signaler och System

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A1.

- f(t): flip about the t axis

1: DC

A2. B is the correct solution.

D cannot be correct, because the convolution of two positive sequences yields a positive output.

A, B, and C differ for the sample x[3], which is obtained in case of maximal overlap between x1[n] and x2[n] $(\sum_{n=0}^{3} x_1[n]x_2[n] = 6$).

A3. The system is instantaneous. Therefore $y[n]$ depends only on $x[n]$. Comparing input and output, it can be observed that

$$
y[n] = \begin{cases} x[n] & \text{if } x[n] \ge 0\\ -x[n] & \text{if } x[n] < 0 \end{cases}
$$

Therefore $y[n] = |x[n]|$

The system is not invertible. For instance, given $y[5] = 3$, we cannot conclude if $x[5] = 3$ or $x[5] =$ −3.

A4. $Y(\omega) = H(\omega)X(\omega)$

 $sin(t)$ contributes to $X(\omega)$ with terms proportional to δ centered at $\omega = \pm 1$ radians/seconds, and we want to filter them out

 $cos(3t)$ contributes to $X(\omega)$ with terms proportional to δ centered at $\omega = \pm 3$ radians/seconds, and we want to maintain them

Therefore, $1 < \omega_1 < 3$, and $\omega_2 > 3$

A5. The terms \cos and \sin are oscillatory and do not contribute to DC. Therefore, $c_0 = 0$.

A6.

$$
f(t) = \mathcal{L}^{-1}\left\{F(s)\right\} = \mathcal{L}^{-1}\left\{\frac{1}{s}\right\} - \frac{3}{2}\mathcal{L}^{-1}\left\{\frac{1}{s+2}\right\} = u(t) - \frac{3}{2}e^{-2t}u(t) = \left(1 - \frac{3}{2}e^{-2t}\right)u(t)
$$

A7. Transfer function: ratio of the system output to input in the Laplace domain, assuming zero initial conditions

$$
s^{2} Y(s) + 10 s Y(s) + 16 Y(s) = X(s)
$$

$$
H(s) = \frac{Y(s)}{X(s)} = \frac{1}{s^{2} + 10 s + 16}
$$

A8. The system has poles in $p = \frac{-2 \pm \sqrt{4-20}}{2}$ $\frac{1}{2}$ = $-1 \pm j2$, i.e. in the LHP. Therefore, the system is stable.

A9.

$$
\sum_{n=0}^{+\infty} 0.4^n = \sum_{n=0}^{+\infty} 0.4^n z^{-n} |_{z=1} = \mathcal{Z} \left\{ 0.4^n u[n] \right\} |_{z=1} = \frac{z}{z - 0.4} |_{z=1} = \frac{1}{0.6} = \frac{10}{6} = \frac{5}{3}
$$

A10.

$$
f[k] = \mathcal{Z}^{-1} \{ F[z] \} = 4 \mathcal{Z}^{-1} \left\{ \frac{1}{z - 3} \right\} + 5 \mathcal{Z}^{-1} \left\{ \frac{1}{z - 2} \right\} = [4 (3)^{k-1} + 5 (2)^{k-1}] u[k-1]
$$

B1. a. Fundamental frequency

$$
\omega_1 = 2 \quad \left[\frac{r}{s}\right], T_1 = \frac{2\pi}{\omega_1} = \frac{2\pi}{2} = \pi \quad \text{[s]}
$$
\n
$$
\omega_2 = 4 \quad \left[\frac{r}{s}\right], T_2 = \frac{2\pi}{\omega_2} = \frac{2\pi}{4} = \frac{\pi}{2} \quad \text{[s]}
$$
\nCommon period $T = m_1 T_1 = m_2 T_2$, with $m_1, m_2 \in \mathbb{Z}$

\n
$$
T = \pi \quad \text{[s]}, m_1 = 1, m_2 = 2
$$
\n
$$
\omega_0 = \frac{2\pi}{\pi} = 2 \quad \left[\frac{r}{s}\right]
$$

b. Coefficients

$$
x(t) = 7 + \sin(2t) + 5\cos\left(4t + \frac{\pi}{3}\right) = 7 + \frac{e^{j2t} - e^{-j2t}}{2j} + 5\frac{e^{j4t}e^{\frac{j\pi}{3}} + e^{-j4t}e^{-\frac{j\pi}{3}}}{2} =
$$

=
$$
7e^{jk\omega_0 t}\Big|_{k=0} + \frac{1}{2j}e^{jk\omega_0 t}\Big|_{k=1} - \frac{1}{2j}e^{jk\omega_0 t}\Big|_{k=-1} + \frac{5}{2}e^{\frac{j\pi}{3}}e^{jk\omega_0 t}\Big|_{k=2} + \frac{5}{2}e^{-\frac{j\pi}{3}}e^{jk\omega_0 t}\Big|_{k=-2}
$$

$$
c_k = \begin{cases} \frac{1}{\pm \frac{1}{2j}} & k = \pm 1\\ \frac{5}{2} e^{\pm \frac{j\pi}{3}} & k = \pm 2\\ 0 & \text{otherwise} \end{cases}
$$

B2. a.
$$
y_a(t) = x_1(t) * x_2(t) = \int_{-\infty}^{+\infty} x_1(\tau) x_2(t - \tau) d\tau
$$

\n $t \le 0$
\n $y_a(t) = 0$
\n $0 \le t \le 1$
\n $y_a(t) = \int_0^t \tau d\tau = \frac{1}{2} \tau^2 \Big|_0^t = \frac{1}{2} t^2$
\n $1 \le t \le 2$
\n $y_a(t) = \int_{-1+t}^1 \tau d\tau = \frac{1}{2} \tau^2 \Big|_{t=1}^1 = \frac{1}{2} - \frac{1}{2} (t - 1)^2 = t - \frac{1}{2} t^2$
\n $t \ge 2$
\n $y_a(t) = 0$
\n**b.**
\n $y_b(t) = x_2(t) * x_1(t) = \int_{-\infty}^{+\infty} x_2(\tau) x_1(t - \tau) d\tau$
\n $t \le 0$
\n $y_b(t) = 0$
\n0 ≤ t ≤ 1

$$
y_b(t) = \int_0^t (t - \tau) \, d\tau = t \, \tau \Big|_0^t - \frac{1}{2} \tau^2 \Big|_0^t = t^2 - \frac{1}{2} t^2 = \frac{1}{2} t^2
$$

$$
1 \le t \le 2
$$

$$
y_b(t) = \int_{-1+t}^1 (t-\tau) \, d\tau = t \tau \Big|_{t-1}^1 - \frac{1}{2} \tau^2 \Big|_{t-1}^1 = t \left(1 - t + 1\right) - \frac{1}{2} + \frac{1}{2} (t-1)^2 = t - \frac{1}{2} t^2
$$

$t\geq 2$

 $y_b(t) = 0$

 $y_a(t) = y_b(t)$, as expected since the convolution is commutative

Note: this figure is not in scale.

Case a: aliasing

Case b: sampling higher than Nyquist rate, we can recover the original signal (even using a practical filter).

B3.