

Transformers, Signaler & System för D3
 SSY080 2011-10-19

1a/ b) Icke linjärt
 $X_3[n]$ släcks ut ($\omega_b = \pm \frac{3\pi}{4}$)

2/ $H(s) = \frac{1}{(s+2)(s+4)}$, $X(s) = \frac{1}{s+2}$
 $Y(s) = \frac{1}{2} \cdot \frac{1}{(s+2)^2} - \frac{1}{4} \cdot \frac{1}{(s+2)} + \frac{1}{4} \cdot \frac{1}{(s+4)}$
 $y(t) = \frac{1}{2} \left((t - \frac{1}{2}) e^{-2t} + \frac{1}{2} e^{-4t} \right) u(t)$

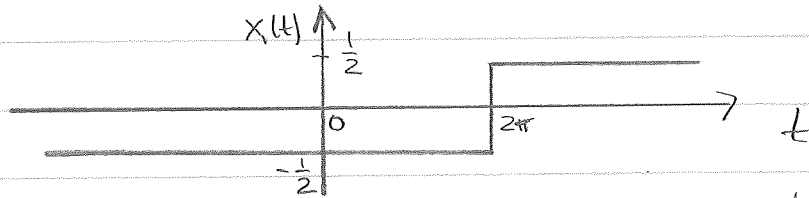
3/ $Y_s(s) = \frac{z}{z-1} \cdot \frac{1}{z} \left(1 + \frac{z-1}{z-\frac{1}{3}} - \frac{z(z-1)}{z-\frac{1}{2}} \right)$
 $Y_s(s) = \frac{z}{z-1} \cdot H(z) \Rightarrow H(z) = \frac{1}{6} \cdot \frac{z}{(z-\frac{1}{2})(z-\frac{1}{3})}$
 $h[n] = \mathcal{Z}^{-1}\{H(z)\} = \left[\left(\frac{1}{2}\right)^n - \left(\frac{1}{3}\right)^n \right] u[n]$

4/a)	k	N-k	b)
E	338	1710	$\Delta f = \frac{f_s}{N} = 0,977 \text{ Hz}$
H	253	1795	
G	201	1847	

5/ (Jämför "Klem-lab")
 $|B_1^Y| = |B_1^X| \cdot |G(j\omega)|_{\omega=\omega_0} = \frac{2}{\pi} \cdot 0,164$
 $|B_2^Y| = |B_2^X| \cdot |G(j\omega)|_{\omega=2\omega_0} = \frac{1}{\pi} \cdot 1$
 $|B_3^Y| = \dots = \frac{2}{3\pi} \cdot 0,287$

2012-01-10

1 a/ $x_1(t) = u(t - 2\pi) - \frac{1}{2}, \forall t$



Är periodisk!

b/ $x_2[n+N] = 3 \cos\left(\frac{(n+N)\pi}{3} - 7\right) =$
 $= 3 \cos\left(\frac{n\pi}{3} - 7 + \frac{N\pi}{3}\right)$

Periodisk om $\frac{N\pi}{3} = 2\pi k \Rightarrow N = 6k$ (Låt $k=1$)

Periodisk med $N=6$

c/ $y[n] = x[7n]$

Insignal	Utsignal
$x_1[n]$	$x_1[7n] = y_1[n]$
$a_1 x_1[n]$	$a_1 x_1[7n] = a_1 y_1[n]$
$x_2[n]$	$x_2[7n] = y_2[n]$
$a_2 x_2[n]$	$a_2 x_2[7n] = a_2 y_2[n]$
$x_3[n] =$ $= a_1 x_1[n] + a_2 x_2[n]$	$y_3[n] = x_3[7n] = a_1 x_1[7n] + a_2 x_2[7n] =$ $= a_1 y_1[n] + a_2 y_2[n]$

Linjärt? Ja!

$$2) \frac{d^2 V_c(t)}{dt^2} + \frac{R}{L} \frac{dV_c(t)}{dt} + \frac{1}{LC} V_c(t) = \frac{1}{LC} V(t)$$

Laplace transf.

$$s^2 V_c(s) + s \frac{R}{L} V_c(s) + \frac{1}{LC} V_c(s) = \frac{1}{LC} V(s)$$

$$a) H(s) = \frac{V_c(s)}{V(s)} = \frac{1/LC}{s^2 + s \frac{R}{L} + \frac{1}{LC}} = \frac{9680}{s^2 + s 139 + 9680}$$

b) Nullstelle: inqa

$$\text{Poler: } s^2 + s 139 + 9680 = 0; s_{1,2} = -\frac{139}{2} \pm \sqrt{\frac{139}{2}^2 - 9680}$$

$$s_{1,2} = -69,5 \pm j69,64$$

c) Kvadratkomplettera

$$H(s) = \frac{9680}{(s + 69,5)^2 + 9680 - 69,5^2} = \frac{9680}{(s + 69,5)^2 + 4849,75} = \frac{9680 \cdot \frac{69,64}{69,64}}{(s + 69,5)^2 + 69,64^2}$$

$$\text{Impulssvar: } h(t) = \mathcal{L}^{-1} \{ H(s) \} =$$

$$= \mathcal{L}^{-1} \left\{ \frac{9680}{69,64} \cdot \frac{69,64}{(s + 69,5)^2 + 69,64^2} \right\} = 139 e^{-69,5t} \sin(69,64t)$$

für $t \geq 0$

$$\begin{aligned}
 3/ \quad h[n] &= \left(\frac{1}{3}\right)^n u[n] + 4\left(\frac{1}{2}\right)^n u[n-1] = \\
 &= \left(\frac{1}{3}\right)^n u[n] + 2 \cdot \left(\frac{1}{2}\right)^{-1} \left(\frac{1}{2}\right)^n u[n-1] = \\
 &= \left(\frac{1}{3}\right)^n u[n] + 2 \left(\frac{1}{2}\right)^{n-1} u[n-1] \quad \text{Z-transformera!}
 \end{aligned}$$

a/

$$\begin{aligned}
 H(z) &= \frac{z}{z - \frac{1}{3}} + 2 \frac{z}{z - \frac{1}{2}} \cdot z^{-1} = \frac{z(z - \frac{1}{2}) + 2(z - \frac{1}{3})}{(z - \frac{1}{3})(z - \frac{1}{2})} = \\
 &= \frac{z^2 + \frac{3}{2}z - \frac{2}{3}}{z^2 - \frac{5}{6}z + \frac{1}{6}} = \frac{1 + \frac{3}{2}z^{-1} - \frac{2}{3}z^{-2}}{1 - \frac{5}{6}z^{-1} + \frac{1}{6}z^{-2}}
 \end{aligned}$$

$$b/ \quad H(z) = Y(z)/X(z)$$

$$Y(z) \left(1 - \frac{5}{6}z^{-1} + \frac{1}{6}z^{-2}\right) = X(z) \left(1 + \frac{3}{2}z^{-1} - \frac{2}{3}z^{-2}\right)$$

Inv. Z-transf.

$$y[n] - \frac{5}{6}y[n-1] + \frac{1}{6}y[n-2] = x[n] + \frac{3}{2}x[n-1] - \frac{2}{3}x[n-2]$$

c/ Stabil? Ja!

Ty kausalt system samt poler till $H(z)$ ligger
 innanför enhetscirkeln $|z| = \frac{1}{2} < 1$
 $|z| = \frac{1}{3} < 1$

4, Samplingsintervall $T = 6,25 \text{ ms}$
 \Rightarrow Samplingsfrekvens $f_s = \frac{1}{T} = 160 \text{ Hz}$

DFT, $N = 64$ punker $X[k]$, $0 \leq k \leq N-1$

Index k motsvarar frekvensen $f_k = \frac{k}{N} \cdot f_s$

Vilket k -värde motsvarar de angivna frekvenserna

f [Hz]	"k" $\frac{f}{f_s} \cdot N$	"N-k"	I figur?
16	6,4	57,6	Ja
32	12,8	51,2	Ja
48	19,2	44,8	Nej
64	25,6	38,4	Ja
80	32	32	Nej <small>= $f_s/2$ ei i figur men tre toppar $\neq f_s/2$ finns</small>
96	38,4	25,6	Ja
112	44,8	19,2	Nej
128	51,2	12,8	Ja
144	57,6	6,4	Ja

De frekvenser med ett inringat " k "-värde kan finnas med i signalen som samplas. Vid närmaste heltalsvärde till " k " har då $|X[k]|$ ett högt värde (en "topp").

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$$5, \quad \hat{X}(t) = A_0 + \sum_{n=1}^{\infty} (A_n \cos(n\omega_0 t) + B_n \sin(n\omega_0 t)) =$$

$$= \frac{2}{\pi} \sum_{n=1}^{\infty} \frac{(-1)^{n+1}}{n} \sin\left(n \frac{\pi}{L} t\right) \quad \Rightarrow \quad A_n = 0, \quad \forall n$$

$$B_n = \frac{2}{\pi} \frac{(-1)^{n+1}}{n}$$

$$\omega_0 = \frac{\pi}{L} = \frac{2\pi}{2L}$$

$$T = 2L = \frac{2\pi}{10} \text{ s}$$

$$B_1 = \frac{2}{\pi}, \quad B_2 = -\frac{1}{\pi}, \quad B_3 = \frac{2}{3\pi}$$

$$\omega_0 = \frac{2\pi}{T} = 10 \text{ r/s}$$

Systemets frekvenssvår

$$G(j\omega) = G(s) \Big|_{s=j\omega} = \frac{400}{(j\omega + 20)^2}$$

$$\text{Amplitudkarakteristik: } |G(j\omega)| = \frac{400}{\omega^2 + 20^2}$$

Amplitud hos utignal (tre lägsta frekvenserna)

$$\boxed{n=1} \quad \omega = \omega_0 = 10$$

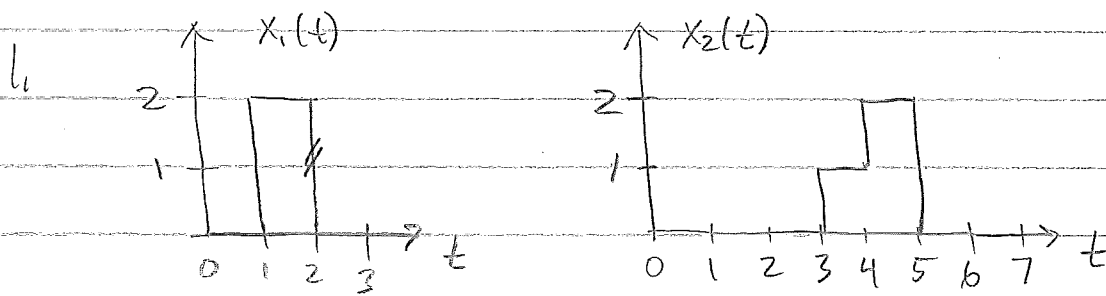
$$B_1^Y = |B_1| \cdot |G(j\omega_0)| = \frac{2}{\pi} \cdot \frac{400}{10^2 + 20^2} = \frac{2}{\pi} \cdot 0,8 = \frac{8}{5\pi} \approx 0,509$$

$$\boxed{n=2} \quad \omega = 2\omega_0 = 20$$

$$B_2^Y = |B_2| \cdot |G(j2\omega_0)| = \frac{1}{\pi} \cdot \frac{400}{20^2 + 20^2} = \frac{1}{\pi} \cdot \frac{1}{2} \approx 0,159$$

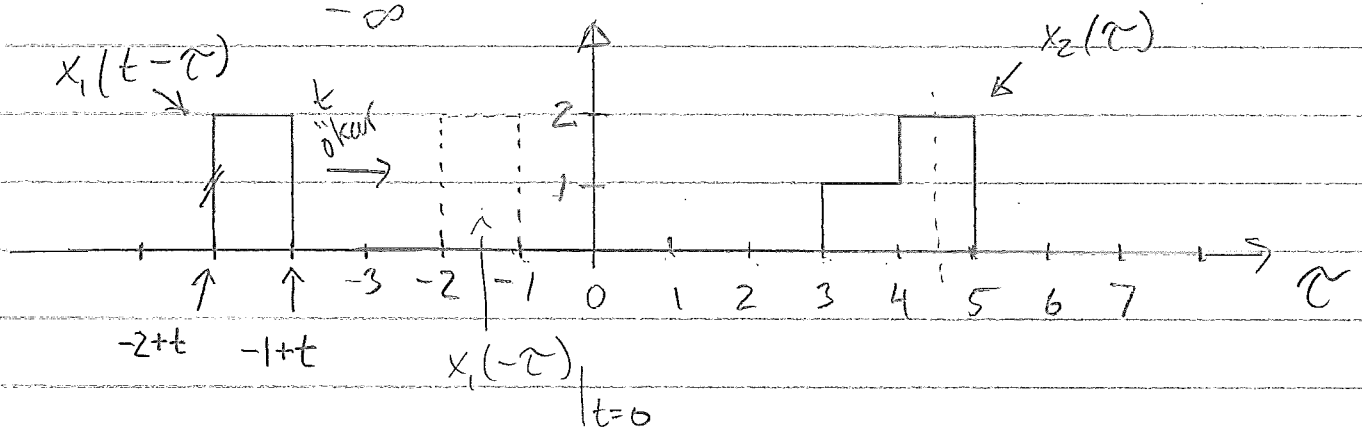
$$\boxed{n=3} \quad \omega = 3\omega_0 = 30$$

$$B_3^Y = |B_3| \cdot |G(j3\omega_0)| = \frac{2}{3\pi} \cdot \frac{400}{20^2 + 30^2} = \frac{2}{3\pi} \cdot \frac{4}{13} \approx 0,065$$



$$y(t) = x_1(t) * x_2(t) = \int_{-\infty}^{\infty} x_1(\tau) x_2(t-\tau) d\tau =$$

$$= \int_{-\infty}^{\infty} x_2(\tau) x_1(t-\tau) d\tau$$



Överlap börjar då $\tau = -1+t = 3 \Rightarrow t = 4$

Överlap avslutas då $\tau = -2+t = 5 \Rightarrow t = 7$

Överlap som ger max $x_1(t) * x_2(t)$: $\tau = -1+t = 5 \Rightarrow t = 6$

$$\text{Max} \int_{-\infty}^{\infty} x_2(\tau) x_1(t-\tau) d\tau \Big|_{t=6} = 2 \cdot 2 \cdot 1 = 4$$

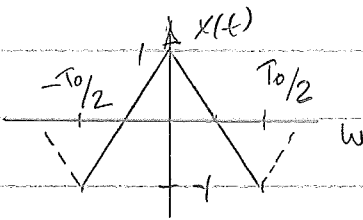
Svar: a) $4 < t < 7$

b) $t = 6$

c) 4

2.

$$H(j\omega) = \frac{j\omega \frac{T_0}{2\pi}}{\left(j\omega \frac{T_0}{2\pi}\right)^2 + j\omega \frac{T_0}{2\pi} + 1}$$



Periodtid : $T = T_0$

Grundvinkelreku. $\omega_0 = \frac{2\pi}{T_0}$

$$H(j\omega) = \frac{j \frac{\omega}{\omega_0}}{1 - \left(\frac{\omega}{\omega_0}\right)^2 + j \frac{\omega}{\omega_0}}$$

$$|H(jk\omega_0)| = \frac{k}{\sqrt{(1-k^2)^2 + k^2}} = \frac{k}{\sqrt{1+k^4-k^2}}$$

$$x(t) = \sum_{k=1,3,5,\dots}^{\infty} \frac{8}{(k\pi)^2} \cos(k\omega_0 t)$$

$$\arg\{H(jk\omega_0)\} = 90^\circ - \arctan\left\{\frac{k}{1-k^2}\right\}$$

Filtret (systemet) H påverkar amplitud och fas hos varje sinusformad signal i $x(t)$ enligt

$$y(t) = \sum_{k=1,3,5,\dots}^{\infty} \frac{8}{(k\pi)^2} \cdot \overbrace{|H(jk\omega_0)|}^{A_k} \cos\left(k\omega_0 t + \overbrace{\arg\{H(jk\omega_0)\}}^{\theta_k}\right)$$

k	ω	A_k	θ_k
1	ω_0	$\frac{8}{\pi^2} \cdot \frac{1}{1} = \frac{8}{\pi^2}$	$90^\circ - \arctan\left(\frac{1}{0}\right) = 0$
3	$3\omega_0$	$\frac{8}{\pi^2} \cdot \frac{1}{3^2} \cdot \frac{3}{\sqrt{1+3^4-3^2}} \approx \frac{8}{\pi^2} \cdot 0,039$	$90^\circ - \arctan\left(\frac{3}{-8}\right) = -69,4^\circ$
5	$5\omega_0$	$\frac{8}{\pi^2} \cdot \frac{1}{5^2} \cdot \frac{5}{\sqrt{1+5^4-5^2}} \approx \frac{8}{\pi^2} \cdot 8,2 \cdot 10^{-3}$	$90^\circ - \arctan\left(\frac{5}{-24}\right) = -78,2^\circ$

$$3. \quad \frac{d^2 y_1(t)}{dt^2} + 3 \frac{dy_1(t)}{dt} + 2y_1(t) = 3x(t) + \frac{dx(t)}{dt}$$

Laplace transformieren!

$$s^2 Y_1(s) + 3s Y_1(s) + 2Y_1(s) = 3X(s) + sX(s)$$

$$H_1(s) = \frac{Y_1(s)}{X(s)} = \frac{s+3}{s^2+3s+2} = \dots = \frac{s+3}{(s+1)(s+2)}$$

$$\mathcal{L}\{\text{"Integrator"}\} = \frac{1}{s}$$

$$Y(s) = X(s) \cdot H_1(s) \cdot \frac{1}{s} \quad \text{Impulssteuer} \Rightarrow x(t) = \delta(t) \xleftrightarrow{\mathcal{L}} X(s) = 1$$

$$Y(s) = \frac{s+3}{s(s+1)(s+2)} = \frac{A}{s} + \frac{B}{s+1} + \frac{C}{s+2} \quad (\text{P.B.U.})$$

$$s+3 = \frac{s^2+3s+2}{s(s+1)(s+2)} + Bs(s+2) + Cs(s+1)$$

$$s^0: \quad 3 = 2A \Rightarrow A = \frac{3}{2}$$

$$s^1: \quad 1 = 3A + 2B + C$$

$$s^2: \quad 0 = A + B + C \Rightarrow C = -A - B$$

$$\left\{ \begin{array}{l} 1 = 3A + 2B - A - B \\ 1 + A - 3A = B \end{array} \right.$$

$$B = 1 + \frac{3}{2} - \frac{9}{2} = -2$$

$$C = -\frac{3}{2} + 2 = \frac{1}{2}$$

$$Y(s) = \frac{3}{2} \cdot \frac{1}{s} - \frac{2}{s+1} + \frac{1}{2} \cdot \frac{1}{s+2}$$

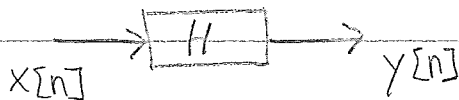
Inv. Laplace quer

$$h(t) = y(t) = \left(\frac{3}{2} - 2e^{-t} + \frac{1}{2}e^{-2t} \right) u(t) \quad [\text{Impulssteuer!}]$$

$$\text{Alt: } h_1(t) = \mathcal{L}^{-1}\{H_1(s)\}$$

$$h(t) = \int_0^t h_1(\tau) d\tau$$

4.



$$y[n] + \frac{1}{3} y[n-1] = 2x[n] + x[n-1], \quad z\text{-transformera!}$$

$$Y(z) \left(1 + \frac{1}{3} z^{-1}\right) = X(z) (2 + z^{-1})$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{2 + z^{-1}}{1 + \frac{1}{3} z^{-1}} = \frac{2z + 1}{z + \frac{1}{3}}$$

$$x[n] = \left(\frac{1}{2}\right)^{n-1} u[n-1] \xleftrightarrow{z} X(z) = \frac{z}{z - \frac{1}{2}} \cdot z^{-1} = \frac{1}{z - \frac{1}{2}}$$

$$Y(z) = H(z)X(z) = \frac{2z + 1}{\left(z + \frac{1}{3}\right)\left(z - \frac{1}{2}\right)} = \frac{2z + 1}{\left(z^2 - \frac{z}{6} - \frac{1}{6}\right)}$$

Teckna $\frac{Y(z)}{z} = \frac{2z + 1}{z\left(z + \frac{1}{3}\right)\left(z - \frac{1}{2}\right)}$ Partialbråksuppdelning

$$\frac{2z + 1}{z\left(z + \frac{1}{3}\right)\left(z - \frac{1}{2}\right)} = \frac{A}{z} + \frac{B}{z + \frac{1}{3}} + \frac{C}{z - \frac{1}{2}}$$

$$2z + 1 = A\left(\frac{z^2}{z} - z \cdot \frac{1}{6} - \frac{1}{6}\right) + Bz\left(z - \frac{1}{2}\right) + Cz\left(z + \frac{1}{3}\right)$$

$$z^0: 1 = A\left(-\frac{1}{6}\right) \Rightarrow A = -6$$

$$z^1: 2 = -\frac{1}{6}A - \frac{1}{2}B + \frac{1}{3}C$$

$$z^2: 0 = A + B + C$$

$$\rightarrow 2 = 1 - \frac{1}{2}B + \frac{1}{3}(6 - B)$$

$$C = -A - B = 6 - B$$

$$A = -6$$

$$B = 1,2$$

$$C = 4,8$$

$$Y(z) = -6 + 1,2 \frac{z}{z + \frac{1}{3}} + 4,8 \frac{z}{z - \frac{1}{2}}$$

Invers z-transform, $y[n] = -6\delta[n] + \left[1,2\left(-\frac{1}{3}\right)^n + 4,8\left(\frac{1}{2}\right)^n\right] u[n]$

5. Samplings frekvens $f_s = 8000$ Hz

$$\omega_s = 2\pi f_s \quad \text{r/s}$$

$$T_s = \frac{2\pi}{\omega_s} = \frac{1}{f_s} \quad \text{sampleintervall}$$

DFT : $N = 256$ punkter $X[k]$, $k = 0, 1, \dots, N-1$

Index k motsvarar frekvensen

$$f_k = \frac{k}{N} \cdot f_s$$

Ur figuren ser vi att två k -värden ger höga $|X[k]|$

$$k = 25 \Rightarrow f_{25} = \frac{25}{256} \cdot 8000 = 781 \text{ Hz}$$

$$k = 43 \Rightarrow f_{43} = \frac{43}{256} \cdot 8000 = 1344 \text{ Hz}$$

$$\text{Frekvensupplösning i DFT: } \Delta f = \frac{f_s}{N} = 31.25 \text{ Hz}$$

Från tabellen ser vi att bästa överensstämmelse fås för knapp "5".

$$\left(\begin{array}{l} 781 \text{ ligger i intervallet } 770 \pm \Delta f \\ 1344 \text{ " " " " " " } 1336 \pm \Delta f \end{array} \right)$$