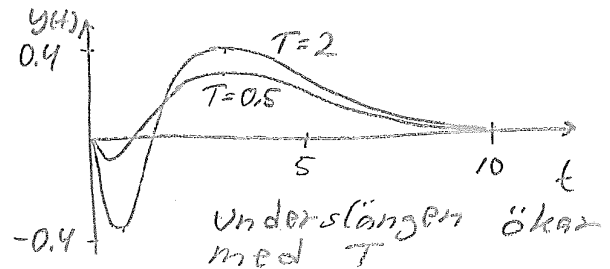


Lösning till tentamen i Reglerteknik 151027

1. a) $G(s) = \frac{1}{(1+s)^3} - T \frac{s}{(1+s)^2}$ $u(t) = \delta(t)$ $U(s) = 1$

$$y(t) = \mathcal{L}^{-1} G(s) = \frac{t^2}{2} e^{-t} - T \frac{d}{dt} \left(\frac{t^2}{2} e^{-t} \right) = \left[(1+T) \frac{t^2}{2} - Tt \right] e^{-t}$$

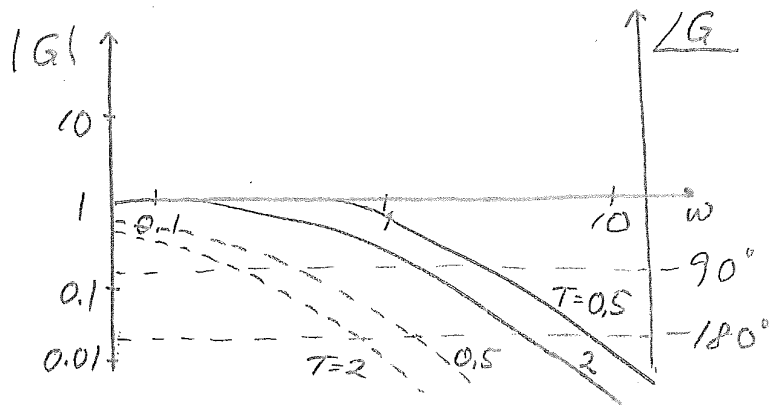
t	$y(t)_{T=0.5}$	$y(t)_{T=2}$
0	0	0
0.5	-0.04	-0.38
3	0.26	0.37
∞	0	0



b) $G(j\omega) = \frac{1 - Tj\omega}{(1 + j\omega)^3}$

$$|G(j\omega)| = \frac{\sqrt{1 + T^2\omega^2}}{\sqrt{1 + \omega^2}^3}$$

$$\angle G(j\omega) = -\arctan T\omega - 3\arctan \omega$$



c) $K = \frac{|G(j\omega_{\pi})|}{|G(0)|} = \begin{cases} 0.31 & T=0.5 \quad (\omega_{\pi} = 1.2) \\ 0.88 & T=2 \quad (\omega_{\pi} = 0.85) \end{cases}$

Då nollstället i $s = 1/T$ närmar sig origo blir processen svårare att reglera, vilket motsvarar ökad undersläng och ökad negativ fäsvridning

2a) $\angle G(j\omega_{G150}) = -4\arctan \omega_{G150} = -150^\circ$

$$\Rightarrow \omega_{G150} = \tan 37.5^\circ = 0.767 \text{ rad/s} \quad \omega_c = 0.4 \omega_{G150} = 0.307 \text{ rad/s}$$

$$\angle G(j\omega_c) + \angle F_{PI}(j\omega_c) = -4\arctan \omega_c - 90^\circ + \arctan \omega_c T_i = -180^\circ + \varphi_m = -130^\circ$$

$$T_i = \frac{1}{\omega_c} \tan(4\arctan \omega_c - 40^\circ) = 1.751$$

$$|G(j\omega_c)| |F_{PI}(j\omega_c)| = \frac{1}{(1 + \omega_c^2)^2} \frac{K_i \sqrt{1 + (\omega_c T_i)^2}}{\omega_c} = 1$$

$$K_i = \frac{\omega_c (1 + \omega_c^2)^2}{\sqrt{1 + (\omega_c T_i)^2}} = 0.324$$

$$\tau_0 = 1/K_i = 3.09$$

$$\tau_u = K_p = K_i T_i = 0.567$$

2 b)

$$\omega_c = 0.6 \cdot \omega_{G150} = 0.6 \cdot 0.767 = 0.460 \text{ rad/s}$$

$$|G(j\omega_c)| = \frac{1}{(1+\omega_c^2)^2} = 0.835 \quad \angle G(j\omega_c) = -4 \arctan \omega_c = -98.8^\circ$$

$$\angle F_{PZD}(j\omega_c) = -180^\circ + \varphi_m - \angle G(j\omega_c) = -180^\circ + 50^\circ + 98.8^\circ = -31.2^\circ$$

$$FS \Rightarrow \omega_c \tau = 0.61 \Rightarrow \tau = 1.326$$

$$K_{\infty} |G(j\omega_c)| = 4.45 \Rightarrow K_{\infty} = K_i \tau \beta = \frac{4.45}{0.835} = 5.33$$

$$K_i = \frac{4.45}{10 \cdot 1.326 \cdot 0.835} = 0.402$$

$$PI: \quad J_0 = 3.09 \quad J_H = 0.567 \quad PID: \quad J_0 = 2.49 \quad J_H = 5.33$$

$$g) \quad \frac{J_{0PID}}{J_{0PI}} = 0.81 \quad \frac{J_{HPID}}{J_{HPI}} = 9.4$$

∴ Nästan 10 gånger högre J_0 och
19% förbättring av J_H

$$3. a) L(s) = \frac{KK_p(s+2)}{s(s-2)}$$

$$G_{ry}(s) = \frac{L(s)}{1+L(s)} = \frac{KK_p s + 2KK_p}{s^2 + (KK_p - 2)s + 2KK_p}$$

Instabilif då $KK_p - 2 \leq 0 \Rightarrow KK_p \leq 2$

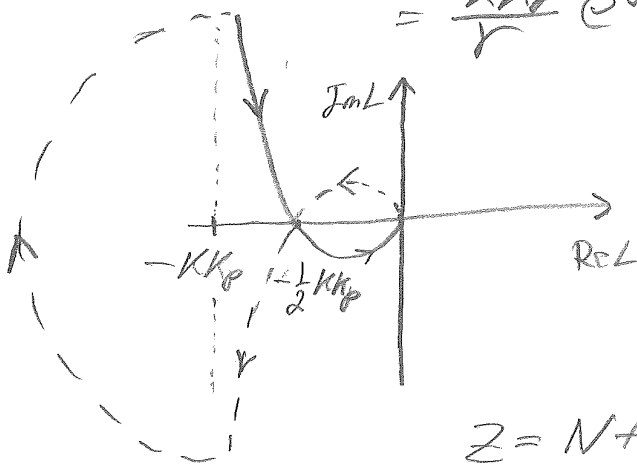
$$b) L(j\omega) = \frac{KK_p(j\omega+2)(-j\omega-2)}{j\omega(j\omega-2)(-j\omega-2)} = -j \frac{KK_p(\omega^2 - 4 - 4j\omega)}{\omega(\omega^2 + 4)}$$

$$= -\frac{4KK_p}{(\omega^2 + 4)} + j \frac{KK_p(4 - \omega^2)}{\omega(4 + \omega^2)} = \begin{cases} -KK_p + j\infty & \omega=0 \\ -\frac{1}{2}KK_p & \omega=2 \\ 0 & \omega=\infty \end{cases}$$

$$L(-j\omega) = \text{Re } L(j\omega) - j \text{Im } L(j\omega)$$

$$L(re^{j\theta}) = -\frac{KK_p}{re^{j\theta}} = \underbrace{e^{j\pi}}_{=-1} \frac{KK_p}{r} e^{-j\theta} = \frac{KK_p}{r} e^{j(\pi-\theta)} =$$

$$= \frac{KK_p}{r} e^{j\varphi} \quad \varphi = \begin{cases} \frac{3\pi}{4} & \theta = -\frac{\pi}{2} \\ \pi & \theta = 0 \\ \frac{\pi}{2} & \theta = \frac{\pi}{2} \end{cases} \quad -\frac{\pi}{2} < \theta < \frac{\pi}{2}$$



$$KK_p < 2 \quad N = 1$$

$$KK_p > 2 \quad N = -1$$

$$Z = N + P = \begin{cases} 1 + 1 = 2 & KK_p \leq 2 \text{ instabilif} \\ -1 + 1 = 0 & KK_p > 2 \text{ stabilif} \end{cases}$$

$$c) L_0(s) = \frac{K_0 K_p (s+2)}{s(s-2)} \quad \text{stabilif då } K_0 K_p > 2$$

$$K_p = \frac{4}{K} \quad (\text{kenominellt värde på } K_0) \Rightarrow$$

$$K_0 \frac{4}{K} = 2 \quad \text{då } K_0 = \frac{K}{2} \text{ dvs instabilitet}$$

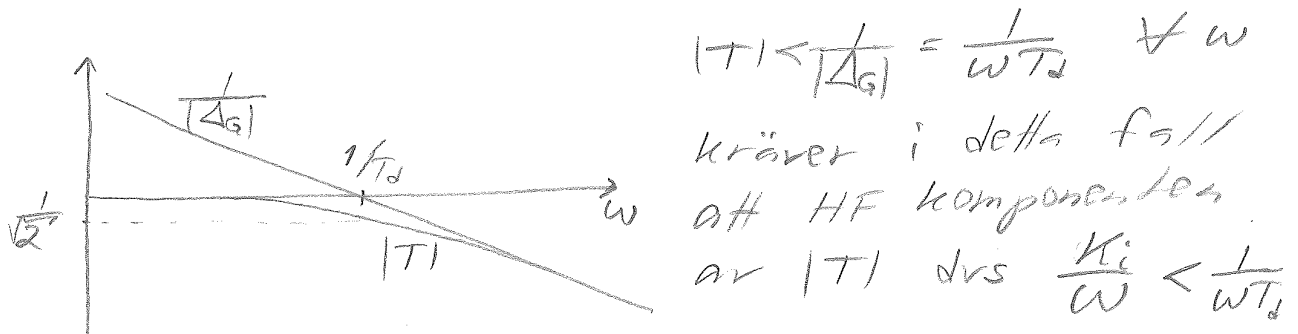
inträffar precis förstärkningen halveras, men stabilitet erhålls för högre processförstärkning.

$$4. \quad G_0(s) = \frac{1 - T_d s}{1 + s s} \quad G(s) = \frac{1}{1 + s s}$$

$$\Delta G(s) = \frac{G_0 - G}{G} = \frac{G_0}{G} - 1 = 1 - T_d s - 1 = -T_d s$$

$$F_{PI}(s) = \frac{K_i (1 + s s)}{s} \Rightarrow L(s) = \frac{K_i}{s} \quad \text{och}$$

$$T(s) = \frac{L(s)}{1 + L(s)} = \frac{K_i}{s + K_i} = \frac{1}{1 + s/K_i}$$



$|T| < \frac{1}{\sqrt{2}} = \frac{1}{\omega T_d} \quad \forall \omega$
kräver i detta fall
att HF komponenten
är $|T|$ dvs $\frac{K_i}{\omega} < \frac{1}{\omega T_d}$

$$|T(j\omega_b)| = \frac{1}{\sqrt{2}} = \frac{1}{\sqrt{1 + \left(\frac{\omega_b}{K_i}\right)^2}} \Rightarrow K_i = \omega_b$$

$$\therefore \omega_b < \frac{1}{T_d}$$

$$5. \quad G_d(z) = z^{-1} \quad F_d(z) = K_p + \frac{K_i z^{-1}}{1 - z^{-1}} = K_p + \frac{K_i}{z - 1} = \frac{K_p z + K_i - K_p}{z - 1}$$

$$a) \quad L_d(z) = G_d(z) F_d(z) = \frac{K_p z + K_i - K_p}{z^2 - z}$$

$$G_{PI}(z) = \frac{L_d(z)}{1 + L_d(z)} = \frac{K_p z + K_i - K_p}{z^2 - (1 - K_p)z + K_i - K_p}$$

$$(z - 0.3)^2 = z^2 - 0.6z + 0.09$$

$$K_p = 0.4 \quad K_i = K_p + 0.09 = 0.49$$

b) Statisk förstärkning: $z = e^{j0} = 1$

$$G_{PI}(1) = \frac{K_p + K_i - K_p}{1 - (1 - K_p) + K_i - K_p} = \frac{K_i}{K_i} = 1$$

\Rightarrow konstante fel undvikas från referenssignal till utsignal

$$6. a) l\ddot{\theta} = g \sin \theta - u \cos \theta$$

$$\dot{w} = \frac{g}{l} \sin \theta - \frac{u}{l} \cos \theta$$

$$\dot{\theta} = w$$

$$\begin{bmatrix} \Delta \dot{\theta} \\ \Delta \dot{w} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{g}{l} \cos \theta_0 + \frac{u_0}{l} \sin \theta_0 & 0 \end{bmatrix}_{\theta_0=0} \begin{bmatrix} \Delta \theta \\ \Delta w \end{bmatrix} + \begin{bmatrix} 0 \\ -\frac{1}{l} \cos \theta_0 \end{bmatrix}_{\theta_0=0} \Delta u$$

Arbetspunkt $\theta_0 = w_0 = u_0 = 0 \Rightarrow$

$$\begin{bmatrix} \dot{\theta} \\ \dot{w} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ g/l & 0 \end{bmatrix} \begin{bmatrix} \theta \\ w \end{bmatrix} + \begin{bmatrix} 0 \\ -1/l \end{bmatrix} u \quad y = \theta = [1 \ 0] \begin{bmatrix} \theta \\ w \end{bmatrix}$$

$$b) l = g = 10 \text{ m/s}^2 \Rightarrow A = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} \quad B = \begin{bmatrix} 0 \\ -0.1 \end{bmatrix}$$

$$S = [B \ AB] = \begin{bmatrix} 0 & -0.1 \\ -0.1 & 0 \end{bmatrix} \quad \det S = -0.01 \neq 0$$

\therefore styrbart system

c) $L_u = [l_\theta \ l_w]$ slutna systemets poler

$$\det(sI_n - A + BL_u) = \det\left(\begin{bmatrix} s & -1 \\ -1 & s \end{bmatrix} + \begin{bmatrix} 0 \\ -0.1 \end{bmatrix} [l_\theta \ l_w]\right) =$$

$$= \det \begin{bmatrix} s & -1 \\ -1 - 0.1 l_\theta & s - 0.1 l_w \end{bmatrix} = s^2 - 0.1 l_w s - 1 - 0.1 l_\theta$$

$$l_w = 0 \Rightarrow s^2 = 1 + 0.1 l_\theta$$

$$\text{poler } s = \begin{cases} \pm \sqrt{1 + 0.1 l_\theta} & l_\theta \geq -10 \\ \pm j \sqrt{0.1 |l_\theta| - 1} & l_\theta < -10 \end{cases}$$

drs alltid pol i HHP eller på imaginäraxeln

$$d) \alpha_c(s) = s^2 + 2\zeta\omega_n s + \omega_n^2 = s^2 + 3.6s + 9$$

$$= s^2 - 0.1 l_w s - 1 - 0.1 l_\theta \Rightarrow l_w = -36 \quad l_\theta = -100$$

$$G_{ro} = C(sI_n - A + BL_u)^{-1} B K_r = [1 \ 0] \begin{bmatrix} s & -1 \\ 9 & s + 3.6 \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ -0.1 \end{bmatrix} K_r =$$

$$= [1 \ 0] \frac{\begin{bmatrix} s + 3.6 & 1 \\ -9 & s \end{bmatrix} \begin{bmatrix} 0 \\ -0.1 \end{bmatrix}}{s^2 + 3.6s + 9} K_r = \frac{-0.1 K_r}{s^2 + 3.6s + 9}$$

$$G_{ro}(0) = \frac{-0.1 K_r}{9} = 1 \Rightarrow K_r = -90$$