

# Lösningar elektronik maj 2013

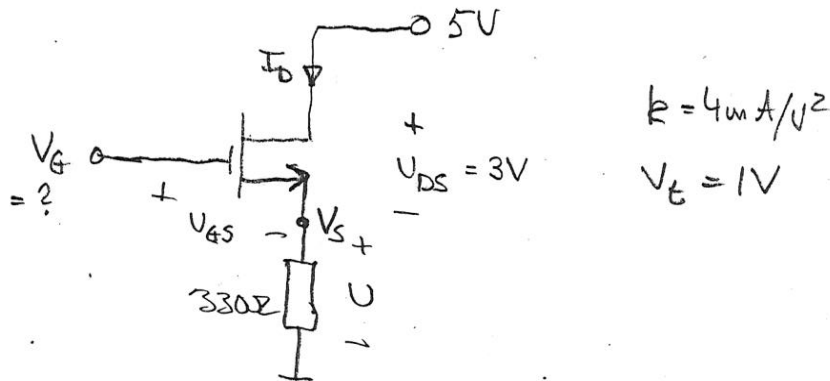
- 1a.
- Massmotstånd lämpliga vid höga frekvenser
  - Metallfilmsmotstånd har inductans och är olämpliga vid höga frekvenser

- 1b. Filterkondensator : • stabil kapacitans  
• små förluster

- Kopplingskondensator : • låg impedans över  
• stort frekvensområde

- 1c Vanliga sekundärbatterier : • Blyackumulator  
• Ni-Cd-ackumulator

1d.



•  $V_G = U + U_{GS}$  ,  $U = 5V - 3V = 2V \Rightarrow I_D = \frac{2V}{330\Omega} = \underline{6,061mA}$

• Antag strömmättnad .  $U_{GS}$  ges av

$I_D = \frac{k}{2} (U_{GS} - V_T)^2 \Rightarrow 6,061mA = \frac{4mA/V^2}{2} (U_{GS} - 1V)^2 \Rightarrow$

$U_{GS} - 1V = \pm 1,74V \Rightarrow U_{GS} = \begin{cases} 2,74V \\ (-0,74V) \text{ strypt} \end{cases}$

Strömmättnad?  $U_{DS} = 3V$   $U_{GS} - V_T = 1,74V$  Ja!

$\Rightarrow V_G = U + U_{GS} = 2V + 2,74V = \underline{4,74V}$

1e lcke-oscillatoriskt: reella poler

$$s^2 + 36s + 24a = 0 \Rightarrow s = -18 \pm \sqrt{18^2 - 24a}$$

reella poler om  $24a \leq 18^2 \Rightarrow \underline{\underline{a \leq 13,5}}$

1F  $I_D = \frac{k}{2} (U_{GS} - V_t)^2 \Rightarrow I_D = \frac{k}{2} (U_{GS} + 4V)^2$

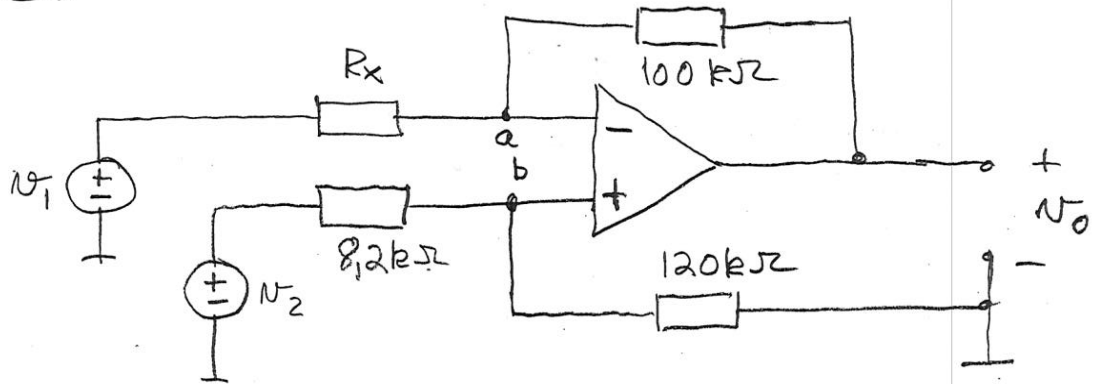
$U_{GS} = 0 \Rightarrow I_D = 8mA \Rightarrow$

$8mA = \frac{k}{2} \cdot (0 + 4V)^2 = k \cdot 8V^2 \Rightarrow \underline{\underline{k = 1mA/V^2}}$

$I_D \geq 0, V_t < 0 \Rightarrow \underline{\underline{N-kanals utarmnings MOSFET}}$

2.

3



Ideal motkopplad OP  $\Rightarrow$

- $U_a = U_b$
- $i_{\text{strömmar}} = 0$

Nodanalys:

$$\frac{U_b - U_2}{8,2 \text{ k}\Omega} + \frac{U_b}{120 \text{ k}\Omega} = 0 \Rightarrow U_b \left( \frac{1}{8,2 \text{ k}\Omega} + \frac{1}{120 \text{ k}\Omega} \right) = \frac{U_2}{8,2 \text{ k}\Omega} \Rightarrow$$

$$U_b \left( 1 + \frac{8,2 \text{ k}\Omega}{120 \text{ k}\Omega} \right) = U_2 \Rightarrow U_b = \frac{120 \text{ k}\Omega}{120 \text{ k}\Omega + 8,2 \text{ k}\Omega} \cdot U_2 = \underline{0,936 \cdot U_2}$$

$$\frac{U_a - U_1}{R_x} + \frac{U_a - U_0}{100 \text{ k}\Omega} = 0 \Rightarrow U_a \left( \frac{1}{R_x} + \frac{1}{100 \text{ k}\Omega} \right) = \frac{U_0}{100 \text{ k}\Omega} + \frac{U_1}{R_x}$$

$$\Rightarrow U_0 = U_a \left( \frac{100 \text{ k}\Omega}{R_x} + 1 \right) - \frac{100 \text{ k}\Omega}{R_x} \cdot U_1 ; U_a = U_b \Rightarrow$$

$$U_0 = 0,936 \cdot \left( \frac{100 \text{ k}\Omega}{R_x} + 1 \right) U_2 - \frac{100 \text{ k}\Omega}{R_x} \cdot U_1 \Rightarrow$$

$$0,936 \cdot \left( \frac{100 \text{ k}\Omega}{R_x} + 1 \right) = \frac{100 \text{ k}\Omega}{R_x} \Rightarrow$$

$$100 \text{ k}\Omega = 0,936 (100 \text{ k}\Omega + R_x) \Rightarrow$$

$$100 \text{ k}\Omega = 93,6 \text{ k}\Omega + 0,936 R_x \Rightarrow \underline{\underline{R_x = 6,83 \text{ k}\Omega}}$$

$$k = \frac{100 \text{ k}\Omega}{6,83 \text{ k}\Omega} = \underline{\underline{14,6}}$$

(4)

$$3. \quad F(s) = \frac{10^{11} \cdot s^2}{(s^2 + 300s + 2 \cdot 10^4)(s + 12000)(s + 25000)}$$

Poler  $s = \begin{cases} -12000 \\ -25000 \end{cases}$

$$s^2 + 300s + 2 \cdot 10^4 = 0 \Rightarrow s = -150 \pm \sqrt{150^2 - 2 \cdot 10^4} =$$

$$= s = -150 \pm 50 = \begin{cases} -200 \\ -100 \end{cases} \Rightarrow$$

$$F(s) = \frac{10^{11} s^2}{(s+100)(s+200)(s+12000)(s+25000)}$$

$$\Rightarrow \omega_{M1} = 100 \text{ rad/s} \quad \omega_{M2} = 200 \text{ rad/s}$$

$$\omega_{\ddot{0}1} = 12000 \text{ rad/s} \quad \omega_{\ddot{0}2} = 25000 \text{ rad/s}$$

$$\Rightarrow \frac{1}{\omega_{\ddot{0}TOT}''} = 1,1 \cdot \sqrt{\frac{1}{12000^2} + \frac{1}{25000^2}} \Rightarrow \omega_{\ddot{0}TOT}'' = \underline{\underline{9,83 \text{ krad/s}}}$$

Stigtid :  $t_r = \frac{2,2}{\omega_{\ddot{0}TOT}''} = \underline{\underline{0,224 \text{ ms}}}$

Relativt pulsfall :

$$P_F = \frac{\Delta t}{t_{TOT}} \cdot 100\% = \Delta t \cdot (\omega_{M1} + \omega_{M2}) \cdot 100\% =$$

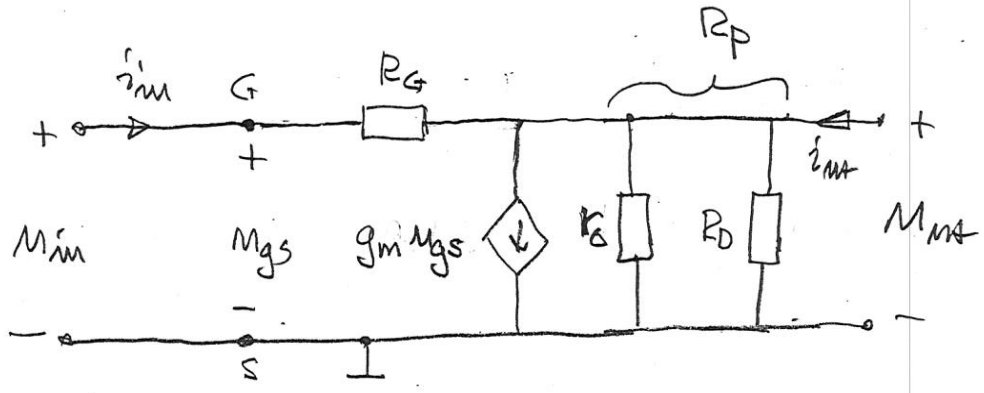
$$= 0,6 \cdot 10^{-3} (100 + 200) \cdot 100\% = \underline{\underline{18\%}}$$

|F|<sub>max</sub> : Tag  $200 \ll s \ll 12000$  :

$$\Rightarrow F(s) \approx \frac{10^{11} s^2}{s \cdot s \cdot 12000 \cdot 25000} = 333 \Rightarrow$$

$$\underline{\underline{|F|_{max} = 333 \text{ ggr} = 50,5 \text{ dB}}}$$

### 4. Small signal scheme



$\frac{U_{out}}{U_{in}}$

- $U_{out} = R_P (i_{in} - g_m U_{GS})$
- $U_{GS} = U_{in}$
- $i_{in} = \frac{U_{in} - U_{out}}{R_G}$

}  $\Rightarrow$

$$U_{out} = R_P \cdot \left( \frac{U_{in} - U_{out}}{R_G} - g_m U_{in} \right) \Rightarrow$$

$$R_G U_{out} = R_P U_{in} - R_P U_{out} - g_m R_P R_G U_{in} \Rightarrow$$

$$\frac{U_{out}}{U_{in}} = \frac{R_P(1 - g_m R_G)}{R_G + R_P} ; R_P = r_o \parallel R_D = 8,246 \text{ k}\Omega$$

$$g_m = 0,25 \text{ mA/V}, R_G = 10 \text{ M}\Omega \Rightarrow$$

$$\frac{U_{out}}{U_{in}} = \underline{\underline{-2,06 \text{ ggr}}}$$

$R_{in}$   $R_{in} = \frac{U_{in}}{i_{in}} ; U_{in} = R_G \cdot i_{in} + R_P (i_{in} - g_m U_{GS}) \Rightarrow$

$$U_{in} = R_G i_{in} + R_P i_{in} - g_m R_P U_{in} \Rightarrow$$

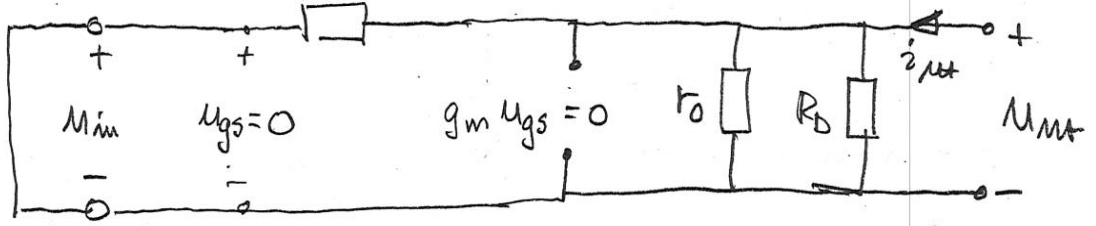
$$R_{in} = \frac{U_{in}}{i_{in}} = \frac{R_G + R_P}{1 + g_m R_P} = \underline{\underline{3,27 \text{ M}\Omega}}$$

$R_{int} \quad R_{int} = \frac{U_{int}}{i_{int}} \quad | \quad U_{in} = 0$

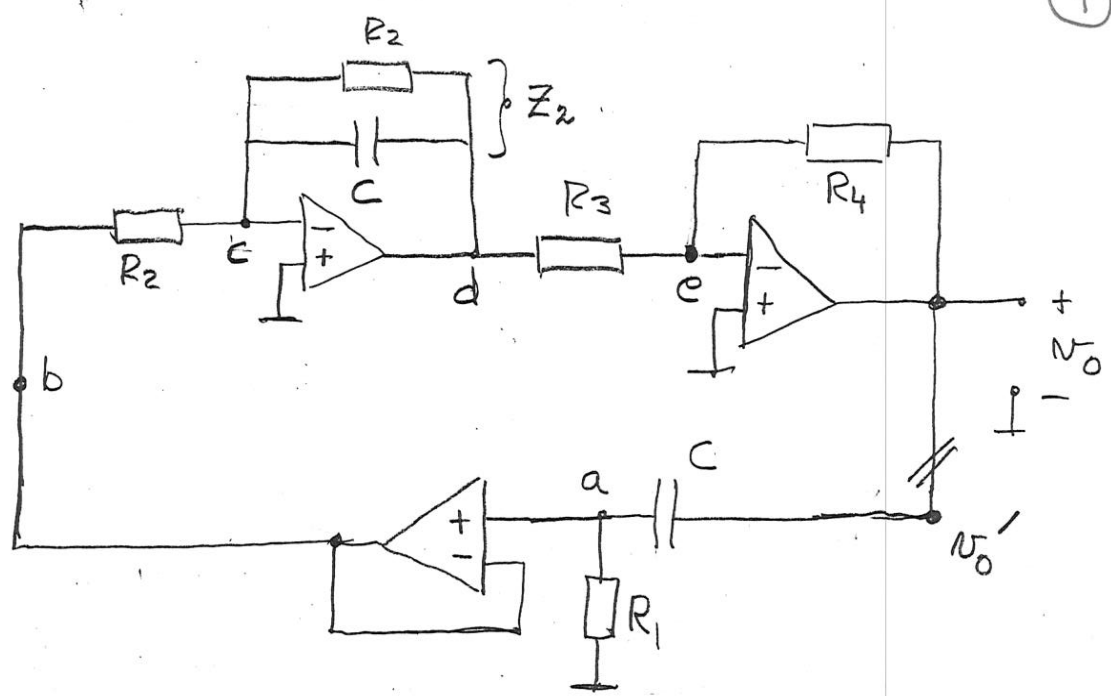
$U_{in} = 0 \Rightarrow U_{gs} = 0 \Rightarrow g_m U_{gs} = 0 \Rightarrow$  (see fig.)

$U_{int} = (R_G // r_o // R_D) \cdot i_{int} \Rightarrow$

$R_{int} = R_G // r_o // R_D = \underline{\underline{8,24 k\Omega}}$



5.



- Ideala motkopplade OP  $\Rightarrow$  inströmmar = 0  
 $V_a = V_b, V_c = V_e = 0$
- Nodanalys för att bestämma  $T(s) = \frac{V_0}{V_0'}$

a.  $\frac{V_a}{R_1} + sC(V_a - V_0') \Rightarrow V_a \left( \frac{1}{R_1} + sC \right) = sC V_0' \Rightarrow$

$V_a = \frac{sR_1C}{1+sR_1C} \cdot V_0' \quad (1)$

$V_b = V_a$  (spänningsföljare)

c.  $V_c = 0 \Rightarrow \frac{0 - V_b}{R_2} + \frac{0 - V_d}{Z_2} = 0 \Rightarrow$

$V_d = -\frac{Z_2}{R_2} \cdot V_b \quad (2)$

$Z_2 = R_2 \parallel \frac{1}{sC} = \frac{R_2 \cdot \frac{1}{sC}}{R_2 + \frac{1}{sC}} = \frac{R_2}{1+sR_2C}$  (2) ger då

$V_d = -\frac{1}{1+sR_2C} \cdot V_b = -\frac{1}{1+sR_2C} \cdot V_a \quad (3)$

e  $U_c = 0 \Rightarrow$

$$\frac{0 - U_d}{R_3} + \frac{0 - U_0}{R_4} = 0 \Rightarrow U_0 = -\frac{R_4}{R_3} \cdot U_d \quad (4)$$

(1), (3), (4) ger

$$U_0 = -\frac{R_4}{R_3} \cdot \left(-\frac{1}{1+sR_2C}\right) \cdot \frac{sR_1C}{1+sR_1C} \cdot U_0' \cdot \frac{sR_4R_1C}{R_3(1+sR_2C)(1+sR_1C)} \cdot U_0'$$

$$\Rightarrow T(s) = \frac{U_0}{U_0'} = \frac{s \cdot \frac{R_4 R_1}{R_3} \cdot C}{s^2 R_1 R_2 C^2 + s(R_1 + R_2)C + 1}$$

$T(j\omega) = 1 \Rightarrow$

$$\frac{j\omega \cdot \frac{R_4 R_1}{R_3} C}{- \omega^2 R_1 R_2 C^2 + 1 + j\omega(R_1 + R_2)C} = 1 \Rightarrow$$

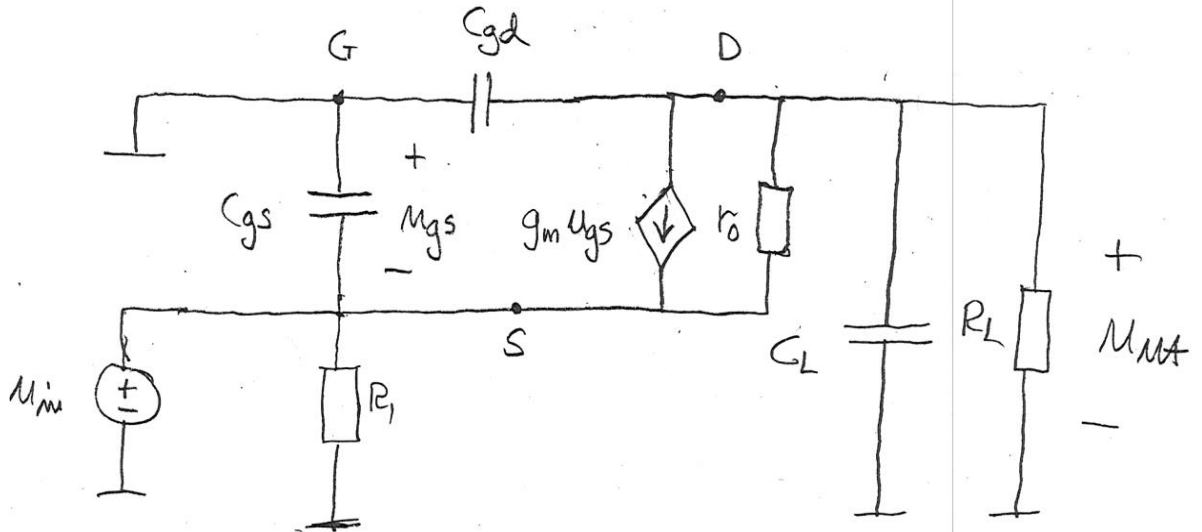
$$\omega = \frac{1}{\sqrt{R_1 \cdot R_2} \cdot C} = \frac{1}{\sqrt{5k \cdot 20k \cdot 15n}} = 6,67 \text{ krad/s} = \underline{\underline{1061 \text{ Hz}}}$$

$$\frac{R_4 R_1}{R_3} \cdot C = (R_1 + R_2)C \Rightarrow R_4 = \frac{R_3 (R_1 + R_2)}{R_1} = \underline{\underline{5k\Omega}}$$

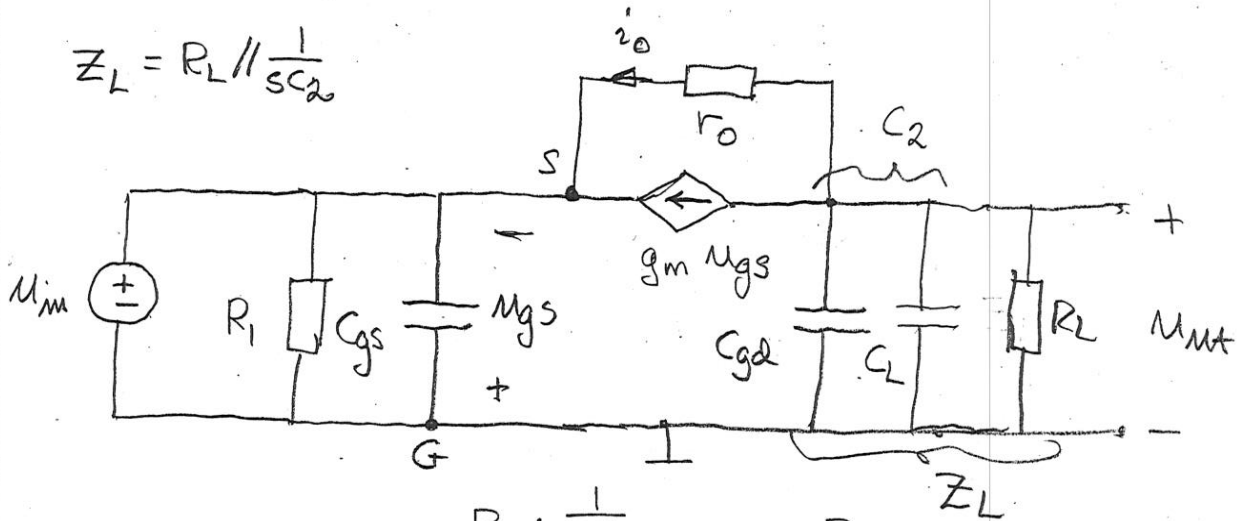


# 6 Småsignalsschema

9



$$Z_L = R_L \parallel \frac{1}{sC_2}$$



$$Z_L = R_L \parallel \frac{1}{sC_2} = \frac{R_L \cdot \frac{1}{sC_2}}{R_L + \frac{1}{sC_2}} = \frac{R_L}{1 + sR_L C_2}$$

$$u_{gs} = -u_{in}$$

$$u_{Mst} = -Z_L \cdot (i_o + g_m u_{gs}) = -Z_L i_o + g_m Z_L \cdot u_{in} \quad \Rightarrow$$

$$i_o = \frac{u_{Mst} - u_{in}}{r_o}$$

$$u_{Mst} = -\frac{Z_L}{r_o} (u_{Mst} - u_{in}) + g_m Z_L \cdot u_{in} \quad \Rightarrow$$

$$r_o u_{Mst} = -Z_L u_{Mst} + Z_L u_{in} + g_m r_o Z_L u_{in} \quad \Rightarrow$$

$$M_{ut}(r_o + Z_L) = Z_L(1 + g_m r_o) M_{in} \Rightarrow$$

$$\frac{M_{ut}}{M_{in}} = (1 + g_m r_o) \cdot \frac{Z_L}{r_o + Z_L}$$

$$r_o + Z_L = r_o + \frac{R_L}{1 + s R_L C_2} = \frac{r_o + R_L + s r_o R_L C_2}{1 + s R_L C_2} =$$

$$= (r_o + R_L) \cdot \frac{1 + s \frac{r_o R_L}{r_o + R_L} C_2}{1 + s R_L C_2} \Rightarrow \left( \frac{r_o R_L}{r_o + R_L} = R_p \right)$$

$$\frac{M_{ut}}{M_{in}} = \frac{1 + g_m r_o}{r_o + R_L} \cdot \frac{\frac{R_L}{1 + s R_L C_2}}{1 + s R_p C_2} =$$

$$= \frac{(1 + g_m r_o) R_L}{r_o + R_L} \cdot \frac{1}{1 + s R_p C_2}$$

$r_o = 20 \text{ k}\Omega$  ,  $C_2 = C_L + C_{gd} = 20 \cdot 10^{-15} \text{ F}$  ,  $R_L = 20 \text{ k}\Omega$

$R_p = r_o \parallel R_L = 10 \text{ k}\Omega$  ,  $g_m = 1,25 \text{ mA/V}$ .

$$\Rightarrow \omega_0 = \frac{1}{R_p C_2} = \frac{1}{10 \text{ k}\Omega \cdot 20 \cdot 10^{-15} \text{ F}} = 5 \text{ Mrad/s} \Rightarrow$$

$f_0 = 796 \text{ kHz}$

Tag  $s \ll 5 \text{ Mrad/s} \Rightarrow$

$$\frac{M_{ut}}{M_{in}} \approx \frac{(1 + g_m r_o) \cdot R_L}{r_o + R_L} \cdot \frac{1}{1 + 0} = \frac{(1 + g_m r_o) \cdot R_L}{r_o + R_L} =$$

$= 13 \text{ gr} = 22,3 \text{ dB}$