

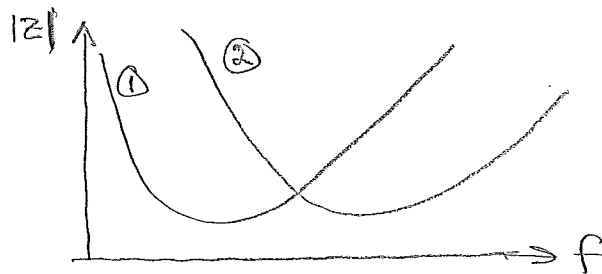
① a. Primärbatterier är icke laddningsbara

- ex. • blysten
- alkaliska
- zink-luft

Sekundärbatterier är laddningsbara

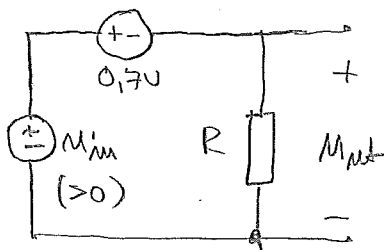
- ex • blyackumulatorer
- Nickel-kadmium-batterier
- Nickel-metallhydrid-batterier
- Litium-jon-batterier

b. Kondensatorer av olika typ har olika frekvensberoende.
 liten impedans över stort frekvensområde
 kräver kombination av olika typer



c. Ekvivalenta scheman

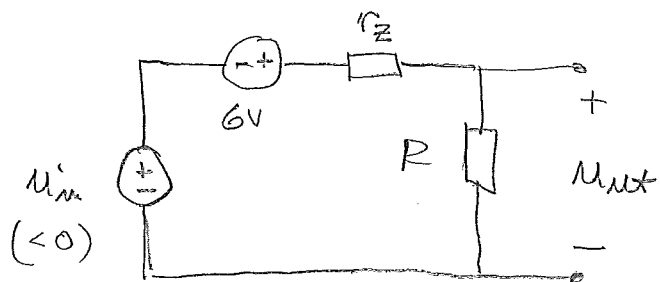
Frånriktning



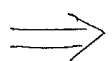
$R = 20 \Omega$

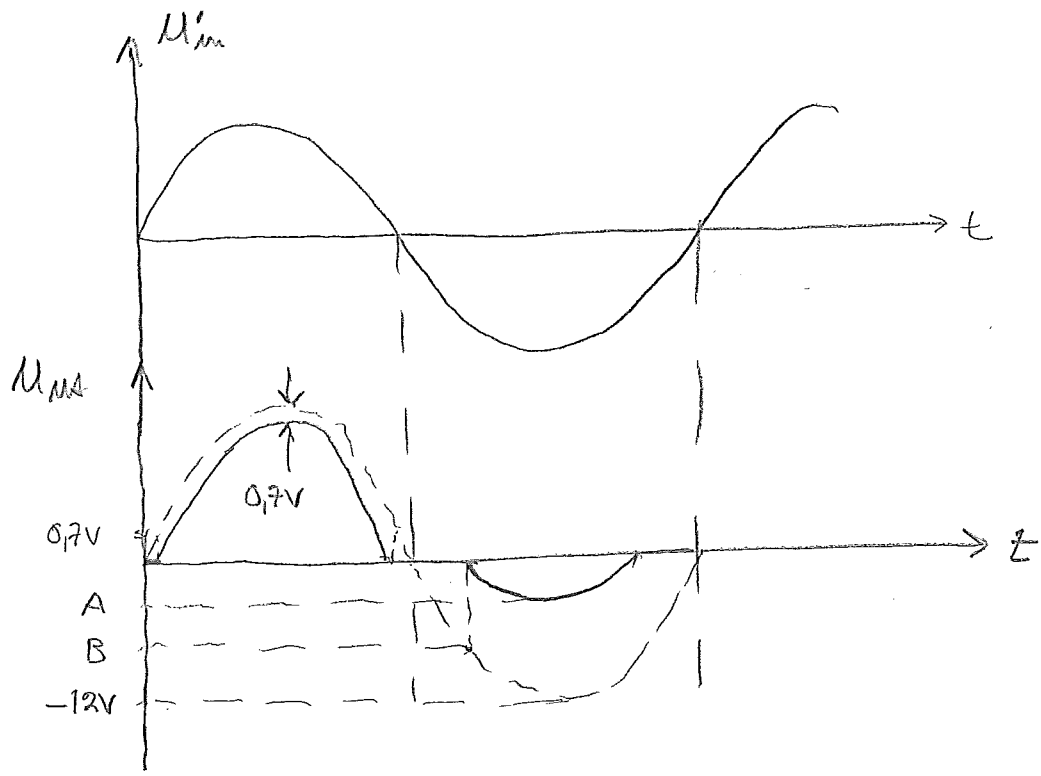
$r_Z = 2 \Omega$

Backriktning



$U_{min} = 12 \cdot \sin \omega t \text{ V}$



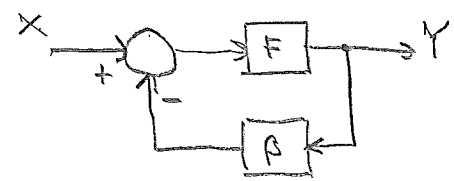


A $M_{\text{max}} = \frac{20 \Omega}{2 \Omega + 20 \Omega} \cdot (-12V + 6V) = \underline{\underline{-5,45V}}$

B $u_{\text{im}} = -6V$ (Zenerdiodeen börjar leda i backriktningen)

1d. Klass A : $\eta_{\text{max}} = \underline{\underline{25\%}}$ Klass B : $\eta_{\text{max}} = \underline{\underline{78,5\%}}$

1e. $F(s) = \frac{A_1 \cdot A_2}{(1 + \frac{s}{\omega_1})(1 + \frac{s}{\omega_2})}$



$$Y = F(s) \cdot (X - \beta \cdot F(s)) \Rightarrow H(s) = \frac{F(s)}{1 + \beta \cdot F(s)} =$$

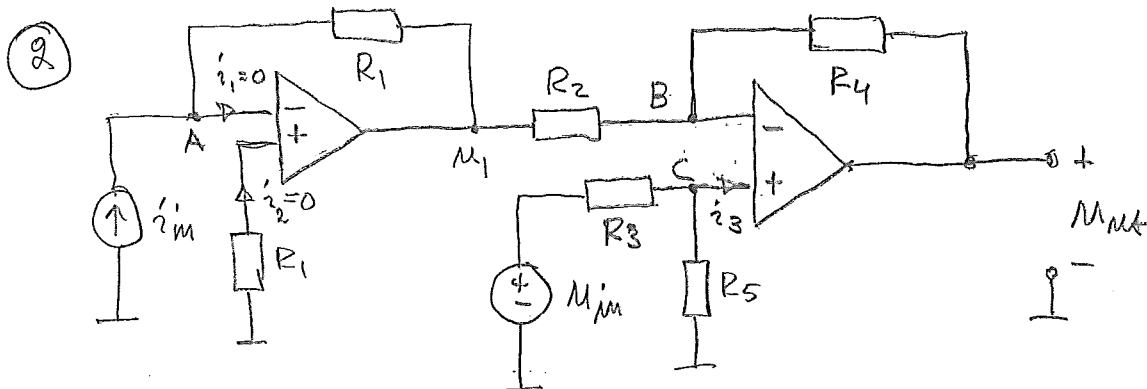
$$= \frac{\frac{A_1 \cdot A_2}{(1 + \frac{s}{\omega_1})(1 + \frac{s}{\omega_2})}}{1 + \beta \cdot \frac{A_1 \cdot A_2}{(1 + \frac{s}{\omega_1})(1 + \frac{s}{\omega_2})}} = \frac{A_1 A_2 \omega_1 \omega_2}{(s + \omega_1)(s + \omega_2) + \beta A_1 A_2 \omega_1 \omega_2}$$

$= \frac{A_1 A_2 \omega_1 \omega_2}{s^2 + (\omega_1 + \omega_2)s + \beta A_1 A_2 \omega_1 \omega_2} \Rightarrow$

$$s = \underbrace{-\frac{\omega_1 + \omega_2}{2}}_{\text{Re}} \pm \underbrace{j \sqrt{\beta A_1 A_2 \omega_1 \omega_2 - \frac{(\omega_1 + \omega_2)^2}{4}}}_{\text{Im}}$$

1F. Z_{in} stor Z_{ut} liten $Z_{in} \gg Z_{ut}$

3



$$U_{ut} = F(i_{in}, U_{in}) = ?$$

Motkopplade ideala OP \Rightarrow instömmar = 0

• lika potentialer på ingångar.

Nodanalys:

$$\underline{A}: i_{in} + \frac{U_A - U_1}{R_1} = 0 ; U_A = R_1 \cdot i_2 = 0 \Rightarrow U_1 = R_1 \cdot i_{in} \quad (1)$$

$$\underline{B}: \frac{U_B - U_1}{R_2} + \frac{U_B - U_{ut}}{R_4} = 0 \Rightarrow U_{ut} = \left(1 + \frac{R_4}{R_2}\right) \cdot U_B - \frac{R_4}{R_2} U_1 \quad (2)$$

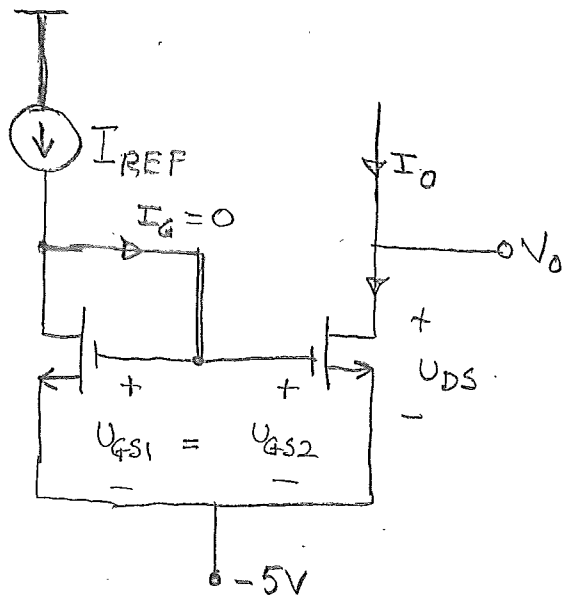
$$i_3 = 0 \Rightarrow U_C = \frac{R_5}{R_5 + R_3} \cdot U_{in} ; U_B = U_C = \frac{R_5}{R_5 + R_3} \cdot U_{in} \quad (3)$$

(1), (2), (3) \Rightarrow

$$U_{ut} = \left(1 + \frac{R_4}{R_2}\right) \cdot \frac{R_5}{R_5 + R_3} \cdot U_{in} - \frac{R_4}{R_2} \cdot R_1 \cdot i_{in}$$

3

4



$$k = 0,08 \text{ mA/V}^2$$

$$V_t = 1 \text{ V}$$

$$V_A = 20 \text{ V}$$

$$I_{REF} = 10 \mu\text{A}$$

Liika resistorer och samma $U_{GS} \Rightarrow$

$$I_0 = I_{REF} = 10 \mu\text{A}$$

Signalvässig utresistans $r_0 = \frac{V_A}{I_0} = \frac{20 \text{ V}}{10 \mu\text{A}} = \underline{\underline{2 \text{ M}\Omega}}$

$U_{GS} = U_{GS1} = U_{GS2}$ ges av $I_{REF} = \frac{k}{2} (U_{GS} - V_t)^2$ vid

strömmättnad. \Rightarrow

$$10 \mu\text{A} = \frac{0,08 \text{ mA/V}^2}{2} (U_{GS} - 1 \text{ V})^2 \Rightarrow U_{GS} = 1 \text{ V} \pm 0,5 \text{ V} = \left. \begin{array}{l} 1,5 \text{ V} \\ 0,5 \text{ V} \end{array} \right\}$$

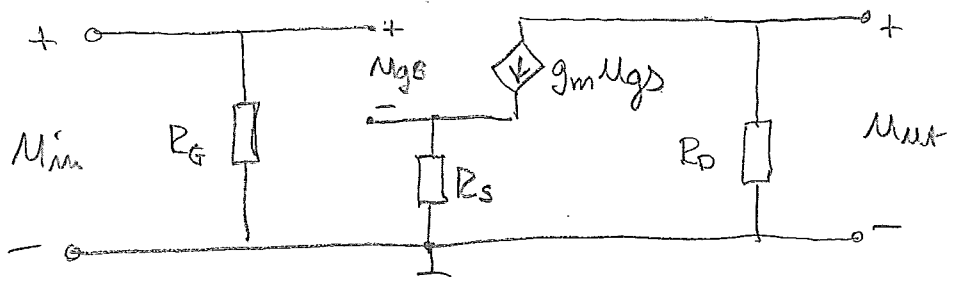
Strömmättnad om $U_{DS} \geq U_{GS} - V_t$,

$$U_{DS} = V_0 - (-5 \text{ V}) = V_0 + 5 \text{ V} \Rightarrow$$

$$V_0 + 5 \text{ V} \geq 1,5 \text{ V} - 1 \text{ V} \Rightarrow \underline{\underline{V_0 \geq -4,5 \text{ V}}}$$

④ AC-analys. Bestäm först ett uttryck för

$\frac{U_{out}}{U_{in}}$ ur småsignalschemat.

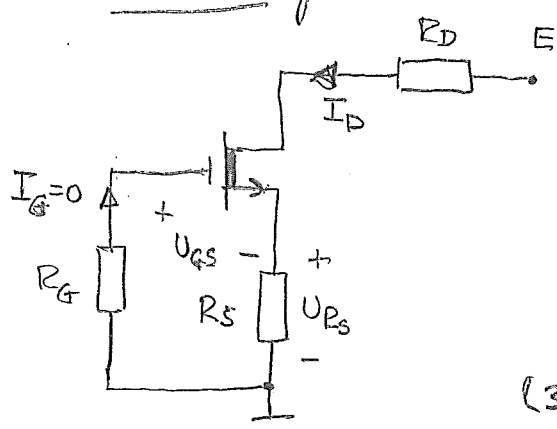


$k = 8 \text{ mA/V}^2$
 $V_t = -1,5 \text{ V}$
 (N-kanal utarmning)
 $R_S = 200 \Omega$
 $\frac{1}{\omega C} \approx 0$

$$\begin{aligned} \bullet U_{in} &= U_{gs} + g_m U_{gs} \cdot R_S = (1 + g_m R_S) \cdot U_{gs} \\ \bullet U_{out} &= -g_m U_{gs} \cdot R_D \end{aligned} \Rightarrow$$

$$\frac{U_{out}}{U_{in}} = - \frac{g_m R_D}{1 + g_m R_S} \quad (1)$$

DC-analys: Bestäm g_m ur $g_m = \sqrt{2k \cdot I_D}$ (2)



$$\begin{aligned} I_G &= 0 \Rightarrow \\ U_{rs} &= -U_{gs} \Rightarrow I_D = \frac{-U_{gs}}{R_S} \quad (3) \end{aligned}$$

$$I_D = \frac{k}{2} (U_{gs} - V_t)^2 \quad (4)$$

$$(3), (4) \Rightarrow - \frac{U_{gs}}{R_S} = \frac{k}{2} (U_{gs} - V_t)^2$$

$$\Rightarrow (U_{gs} - V_t)^2 + \frac{2}{k \cdot R_S} \cdot U_{gs} = 0$$

$$\Rightarrow U_{gs}^2 + \left(\frac{2}{k R_S} - 2V_t\right) \cdot U_{gs} + V_t^2 = 0 \Rightarrow \text{(utan enheter)}$$

$$U_{gs}^2 + 4,25 U_{gs} + 2,25 = 0 \Rightarrow U_{gs} = \begin{cases} (-3,63 \text{ V}) & \text{(strykt)} \\ -0,625 \text{ V} \end{cases}$$

$$\Rightarrow I_D = - \frac{-0,625 \text{ V}}{200 \Omega} = 3,125 \text{ mA} \cdot (2) \Rightarrow g_m = 7 \text{ mA/V}$$

$$(1) \Rightarrow \frac{7 \text{ mA/V} \cdot R_D}{1 + 7 \text{ mA/V} \cdot 200 \Omega} = 3 \Rightarrow R_D = \underline{\underline{1,03 \text{ k}\Omega}}$$

5) Stegen belastar ej verandras \Rightarrow

6

$$F_{TOT}(s) = F_1(s) \cdot F_2(s); \text{ s\u00e5 att}$$

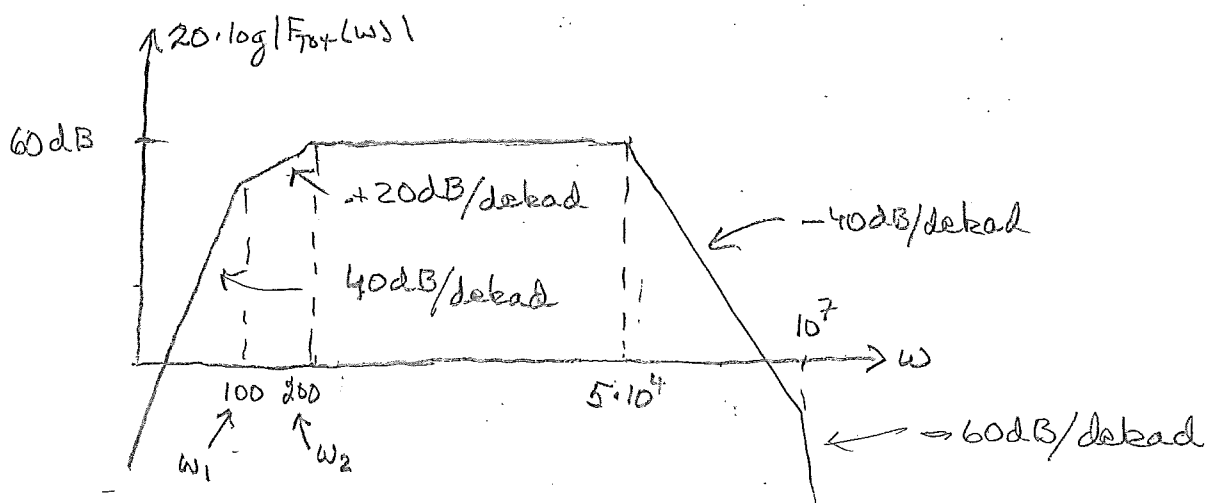
$$F_{TOT}(s) = \frac{-2 \cdot 10^{-3} \cdot s^2}{\left(1 + \frac{s}{100}\right) \left(1 + \frac{s}{200}\right) \left(1 + \frac{s}{3000}\right) \left(1 + \frac{s}{5 \cdot 10^4}\right)} \cdot \frac{25 \cdot \left(1 + \frac{s}{3000}\right)}{\left(1 + \frac{s}{5 \cdot 10^4}\right) \left(1 + \frac{s}{10^7}\right)} =$$

$$= \frac{5 \cdot 10^{-2} \cdot s^2}{\left(1 + \frac{s}{100}\right) \left(1 + \frac{s}{200}\right) \left(1 + \frac{s}{5 \cdot 10^4}\right)^2 \left(1 + \frac{s}{10^7}\right)}$$

$|F_{TOT}|_{max}$ ges av $200 \ll \omega \ll 5 \cdot 10^4$. D\u00e5 g\u00e4ller

$$|F_{TOT}(\omega)| \approx \frac{5 \cdot 10^{-2} s^2}{\frac{s}{100} \cdot \frac{s}{200} \cdot 1 \cdot 1} = 5 \cdot 10^{-2} \cdot 100 \cdot 200 = 10^3 \text{ ggr} = \underline{\underline{60 \text{ dB}}}$$

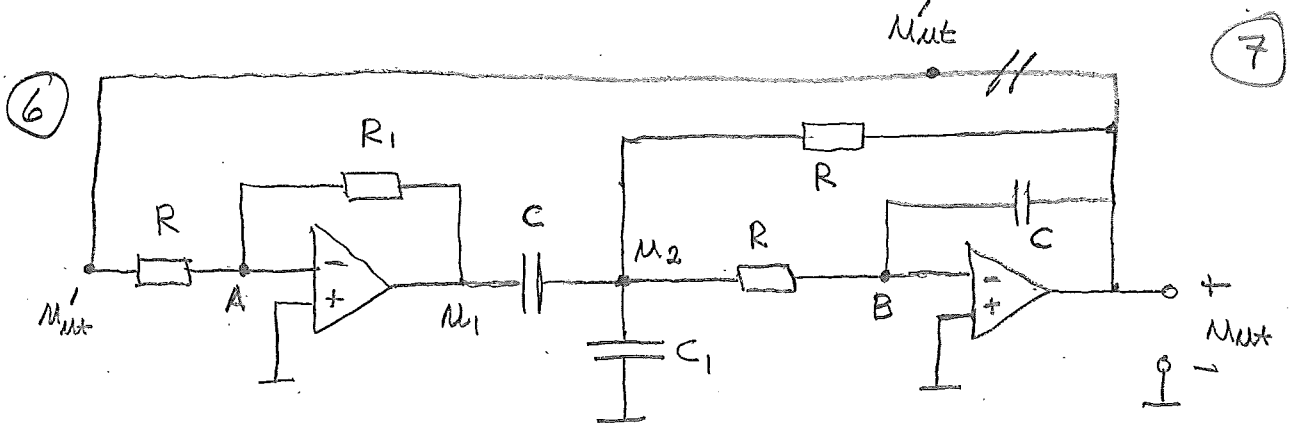
Bode-diagramm f\u00f6r $F_{TOT}(\omega)$:



$$\text{Dubbelpol} \Rightarrow \omega_0 = 5 \cdot 10^4 \cdot \sqrt{2^{1/2} - 1} = 32,18 \text{ krad/s} \Rightarrow$$

$$\underline{\underline{F_0 = 5,12 \text{ kHz}}}$$

$$P_f = (\omega_1 + \omega_2) \cdot \Delta t \cdot 100\% = (100 + 200) \cdot 0,5 \cdot 10^{-3} \cdot 100\% = \underline{\underline{15\%}}$$



Ideala motkopplade OP \Rightarrow instömmar = 0, lika potential på ingångarna.

Bestäm $T(s) = \frac{M_{int}}{M'_{int}}$ och lös ekvationen $T(j\omega) = 1$.

Nodanalys:

A $\frac{U_A - M'_{int}}{R} + \frac{U_A - M_1}{R_1} = 0; U_A = 0 \Rightarrow M_1 = -\frac{R_1}{R} \cdot M'_{int} \quad (1)$

B $\frac{U_B - M_2}{R} + \frac{U_B - M_{int}}{1/sC} = 0; U_B = 0 \Rightarrow M_2 = -sRC \cdot M_{int} \quad (2)$

M_2 $\frac{M_2 - M_1}{1/sC} + \frac{M_2}{1/sC_1} + \frac{M_2 - U_B}{R} + \frac{M_2 - M_{int}}{R} = 0; U_B = 0 \Rightarrow$

$$M_2(sRC + sRC_1 + 2) - sRC \cdot M_1 = M_{int} \quad (3)$$

(1) och (2) i (3) ger

$$-sRC(sRC + sRC_1 + 2) \cdot M_{int} + sRC \cdot \frac{R_1}{R} \cdot M_{int} = M_{int} \Rightarrow$$

$$M_{int} (1 + s^2 R^2 C(C + C_1) + 2sRC) = sR_1 C \cdot M_{int} \Rightarrow$$

$$T(s) = \frac{sR_1 C}{1 + s^2 R^2 C(C + C_1) + 2sRC} \Rightarrow T(j\omega) = \frac{j\omega R_1 C}{1 - \omega^2 R^2 C(C + C_1) + j\omega 2RC}$$

$$T(j\omega) = 1 \Rightarrow 1 - \omega^2 R^2 C(C + C_1) = 0 \Rightarrow$$

$$C_1 = \frac{1}{\omega^2 R^2 C} - C = \underline{\underline{77 \text{ nF}}}$$

$$\frac{j\omega R_1 C}{j\omega 2RC} = 1 \Rightarrow R_1 = 2R = \underline{\underline{40 \text{ k}\Omega}}$$